

Elastic PMLs for Resonator Anchor Loss Simulation

David Bindel¹ and Sanjay Govindjee²

¹ Department of Electrical Engineering and Computer Science

² Department of Civil Engineering
University of California at Berkeley

Contributors

- Tsuyoshi Koyama – PhD Student, Civil Engineering
- Wei He – PhD Student, Civil Engineering
- Emmanuel Quévy – Postdoc, Electrical Engineering
- Roger Howe – Professor, Electrical Engineering
- James Demmel – Professor, Computer Science

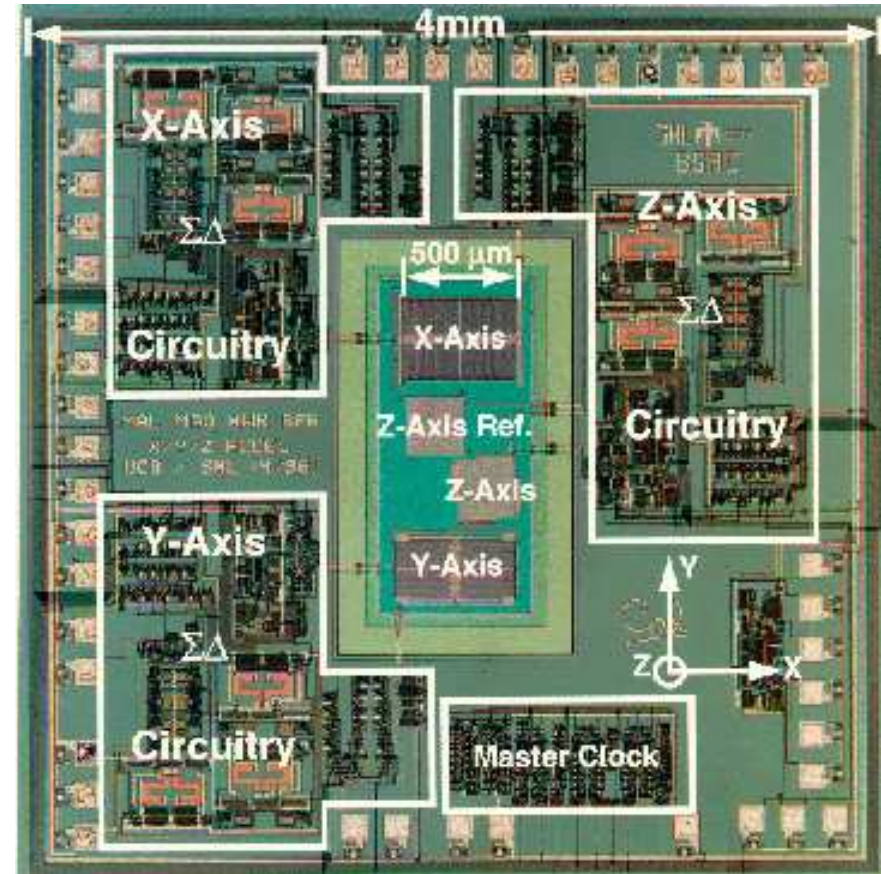
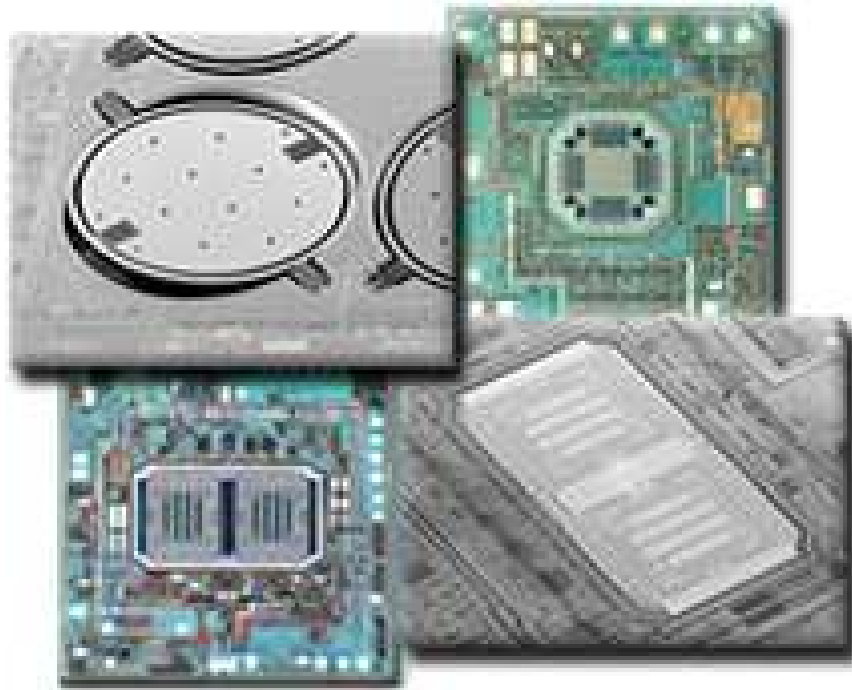
Outline

- Electromechanical resonators and RF MEMS
- Damping and quality of resonance
- Anchor losses and Perfectly Matched Layers
- Analysis of the discretized PMLs
- Complex symmetry and structured model reduction
- Analysis of a disk resonator
- Conclusions

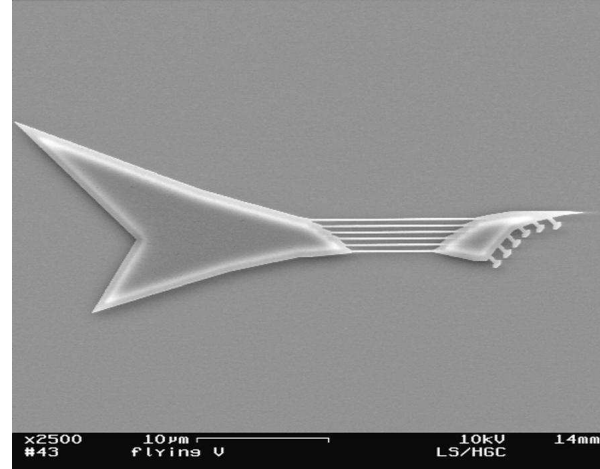
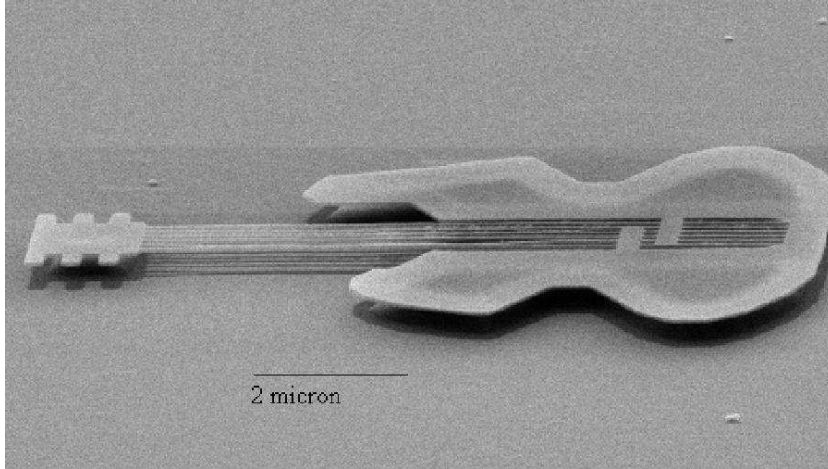
Outline

- Electromechanical resonators and RF MEMS
- Damping and quality of resonance
- Anchor losses and Perfectly Matched Layers
- Analysis of the discretized PMLs
- Complex symmetry and structured model reduction
- Analysis of a disk resonator
- Conclusions

How many MEMS?



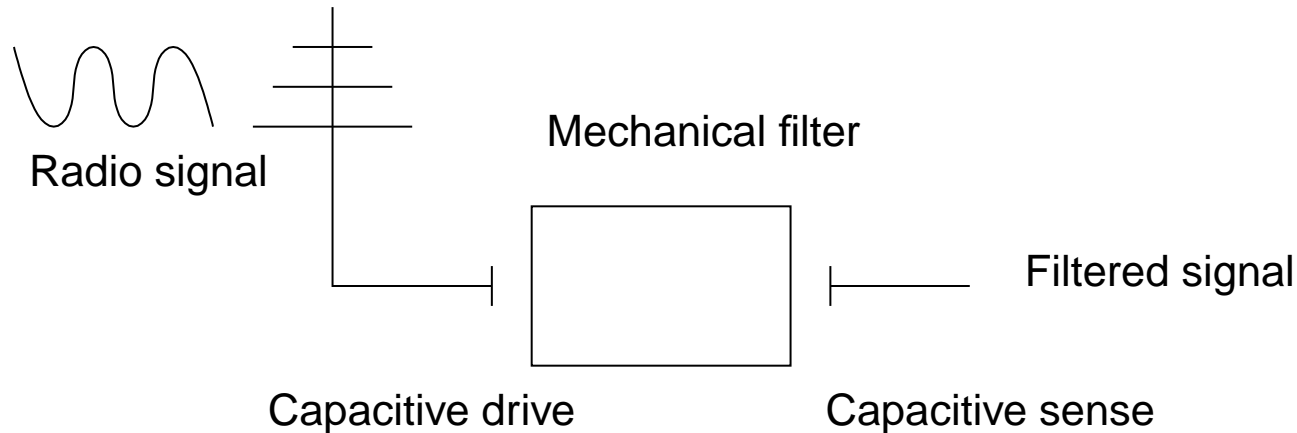
Why resonant MEMS?



Microguitars from Cornell University (1997 and 2003)

- Sensing elements (inertial, chemical)
- Frequency references
- Filter elements
- Neural networks
- Really high-pitch guitars

Micromechanical filters



- Mechanical high-frequency (high MHz-GHz) filter
 - Your cell phone is mechanical!
- Advantage over quartz surface acoustic wave filters
 - Integrated into chip
 - Low power

Success \implies “Calling Dick Tracy!”

Designing transfer functions

Time domain:

$$\begin{aligned}Mu'' + Cu' + Ku &= b\phi(t) \\ y(t) &= p^T u\end{aligned}$$

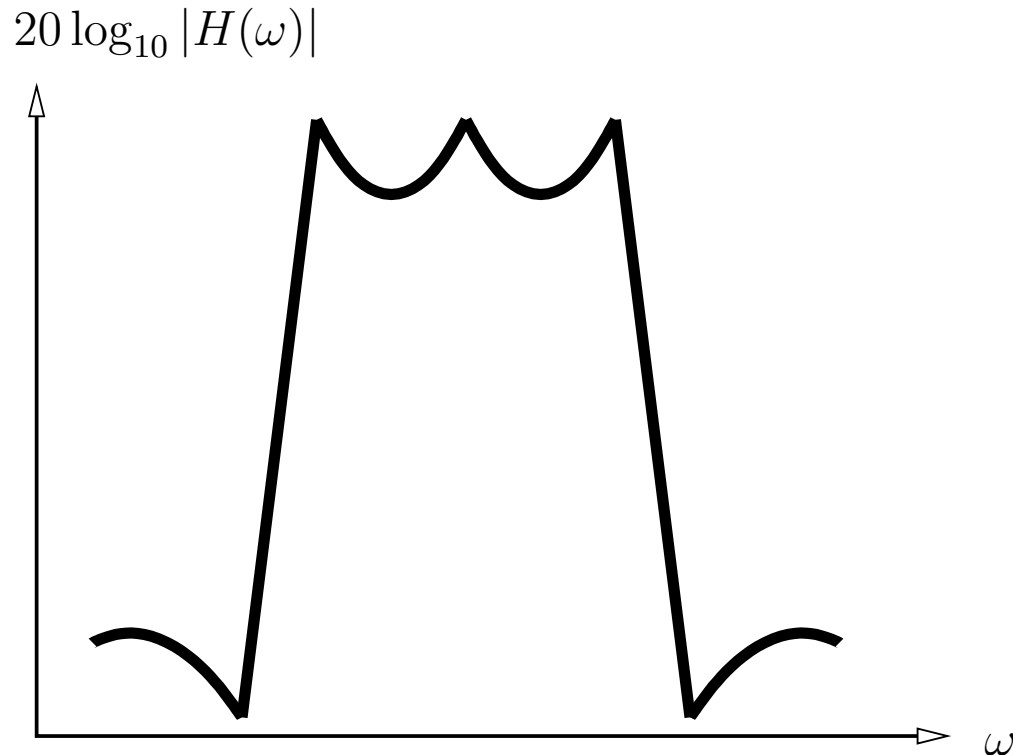
Frequency domain:

$$\begin{aligned}-\omega^2 M\hat{u} + i\omega C\hat{u} + K\hat{u} &= b\hat{\phi}(\omega) \\ \hat{y}(\omega) &= p^T \hat{u}\end{aligned}$$

Transfer function:

$$\begin{aligned}H(\omega) &= p^T (-\omega^2 M + i\omega C + K)^{-1} b \\ \hat{y}(\omega) &= H(\omega)\hat{\phi}(\omega)\end{aligned}$$

Damping and filters



- Want “sharp” poles for narrowband filters
- \implies Want to minimize damping
 - Electronic filters have too much
 - Understanding of damping in MEMS is lacking

Damping and Q

- Designers want high *quality of resonance* (Q)
 - Dimensionless damping in a one-dof system:

$$\frac{d^2u}{dt^2} + Q^{-1} \frac{du}{dt} + u = F(t)$$

- For a resonant mode with frequency $\omega \in \mathbb{C}$:

$$Q := \frac{|\omega|}{2 \operatorname{Im}(\omega)} = \frac{\text{Stored energy}}{\text{Energy loss per radian}}$$

Sources of damping

- Fluid damping
 - Air is a viscous fluid ($Re \ll 1$)
 - Can operate in a vacuum
 - Shown not to dominate in many RF designs
- Material losses
 - Low intrinsic losses in silicon, diamond, germanium
 - Terrible material losses in metals
- Thermoelastic damping
 - Volume changes induce temperature change
 - Diffusion of heat leads to mechanical loss
- Anchor loss
 - Elastic waves radiate from structure

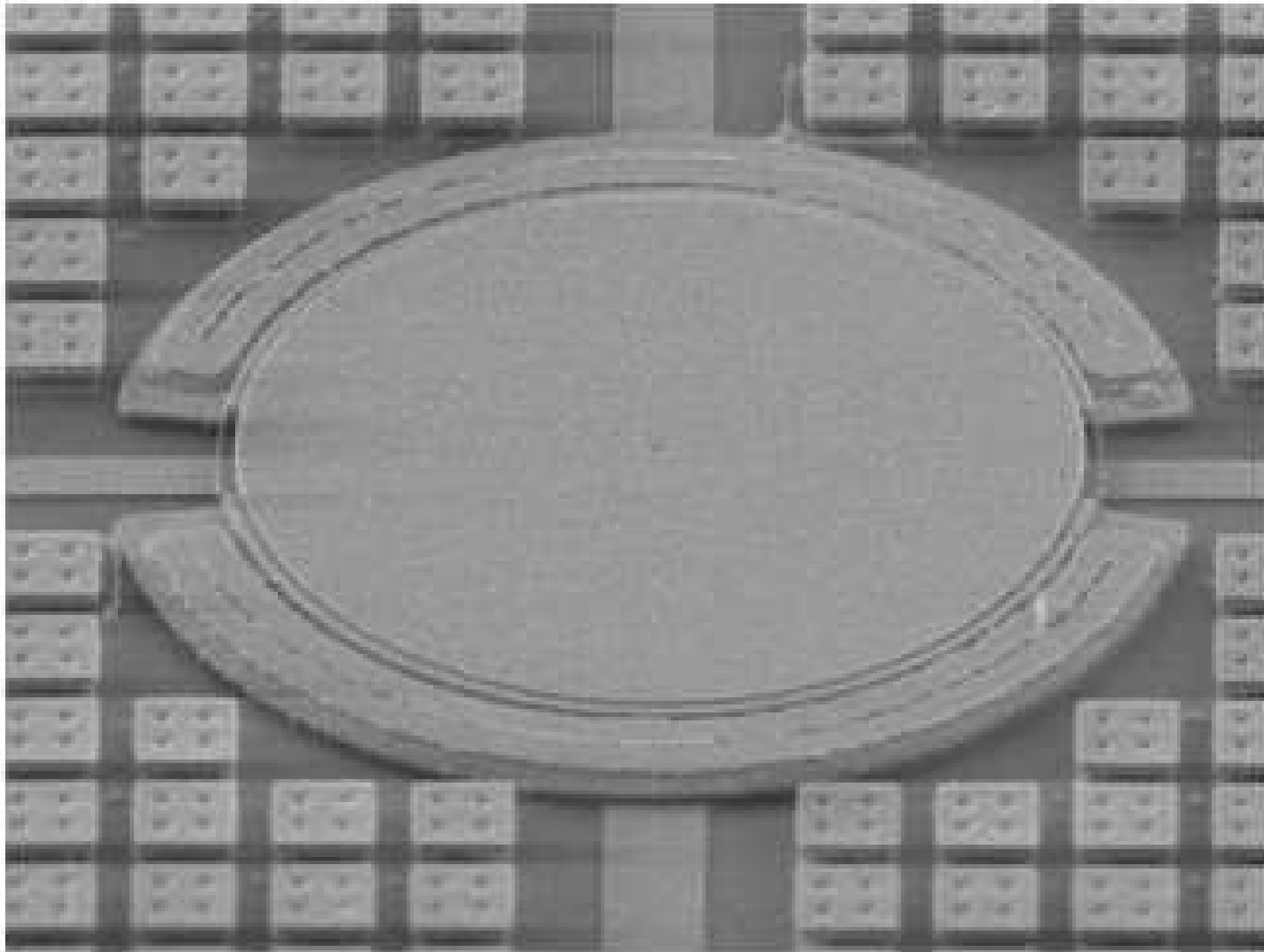
Sources of damping

- Fluid damping
 - Air is a viscous fluid ($Re \ll 1$)
 - Can operate in a vacuum
 - Shown not to dominate in many RF designs
- Material losses
 - Low intrinsic losses in silicon, diamond, germanium
 - Terrible material losses in metals
- Thermoelastic damping
 - Volume changes induce temperature change
 - Diffusion of heat leads to mechanical loss
- Anchor loss
 - Elastic waves radiate from structure

Outline

- Electromechanical resonators and RF MEMS
- Damping and quality of resonance
- Anchor losses and Perfectly Matched Layers
 - Anchor losses and infinite domains
 - Idea of the perfectly matched layer
 - Elastic PMLs and finite elements
- Analysis of the discretized PMLs
- Complex symmetry and structured model reduction
- Analysis of a disk resonator
- Conclusions

Example: Disk resonator



SiGe disk resonators built by E. Quévy

Substrate model

Goal: Understand energy loss in disk resonator

- Dominant loss is elastic radiation from anchor
- Resonator size \ll substrate size
 - Substrate appears semi-infinite
- Possible semi-infinite models
 - Matched asymptotic modes
 - Dirichlet-to-Neumann maps
 - Boundary dampers
 - Higher-order local ABCs
 - Perfectly matched layers

Perfectly matched layers

- Apply a complex coordinate transformation
- Generates a non-physical absorbing layer
- No impedance mismatch between the computational domain and the absorbing layer
- Idea works with general linear wave equations
 - First applied to Maxwell's equations (Bereng er 95)
 - Similar idea earlier in quantum mechanics (*exterior complex scaling*, Simon 79)
 - Applies to elasticity in standard FEM framework (Basu and Chopra, 2003)

1-D model problem

- Domain: $x \in [0, \infty)$

- Governing eq:

$$\frac{\partial^2 u}{\partial x^2} - \frac{1}{c^2} \frac{\partial^2 u}{\partial t^2} = 0$$

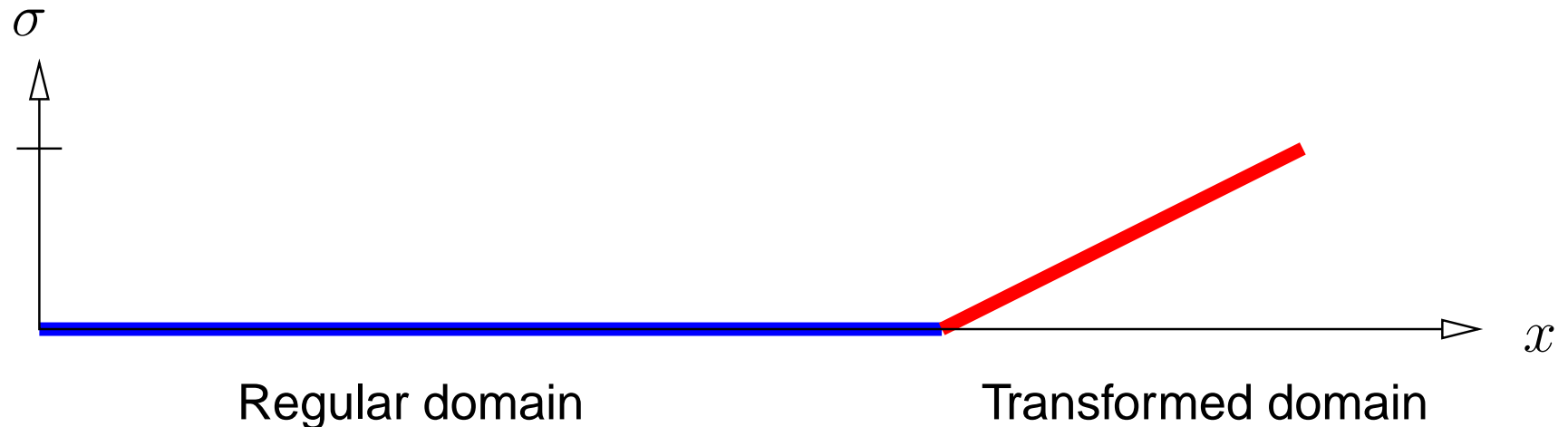
- Fourier transform:

$$\frac{d^2 \hat{u}}{dx^2} + k^2 \hat{u} = 0$$

- Solution:

$$\hat{u} = c_{\text{out}} e^{-ikx} + c_{\text{in}} e^{ikx}$$

1-D model problem with PML

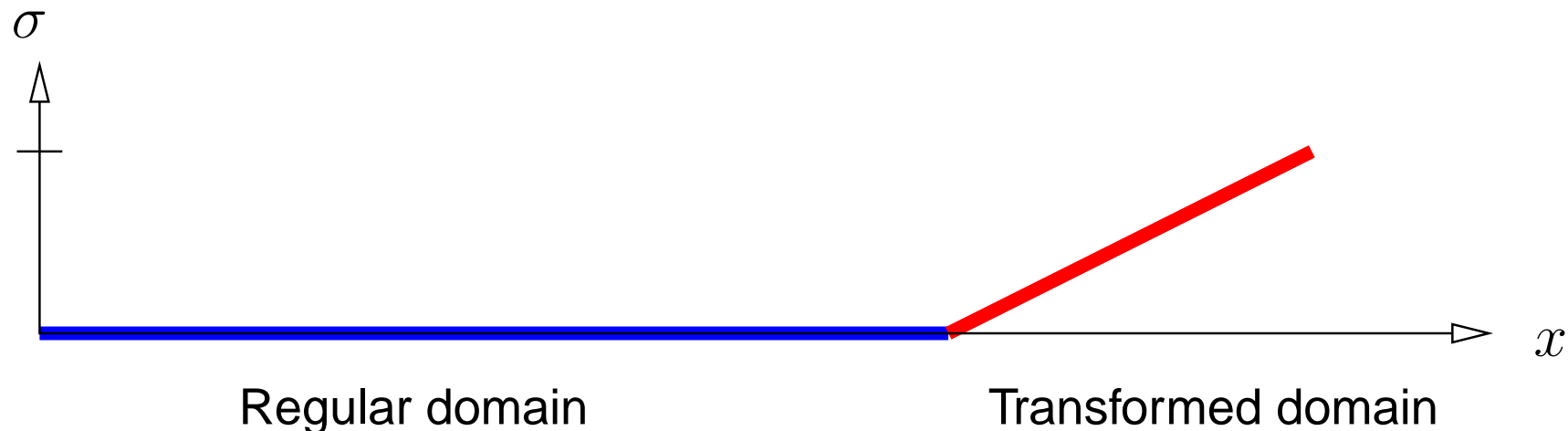


$$\frac{d\tilde{x}}{dx} = \lambda(x) \text{ where } \lambda(s) = 1 - i\sigma(s)$$

$$\frac{d^2\hat{u}}{d\tilde{x}^2} + k^2\hat{u} = 0$$

$$\hat{u} = c_{\text{out}}e^{-ik\tilde{x}} + c_{\text{in}}e^{ik\tilde{x}}$$

1-D model problem with PML

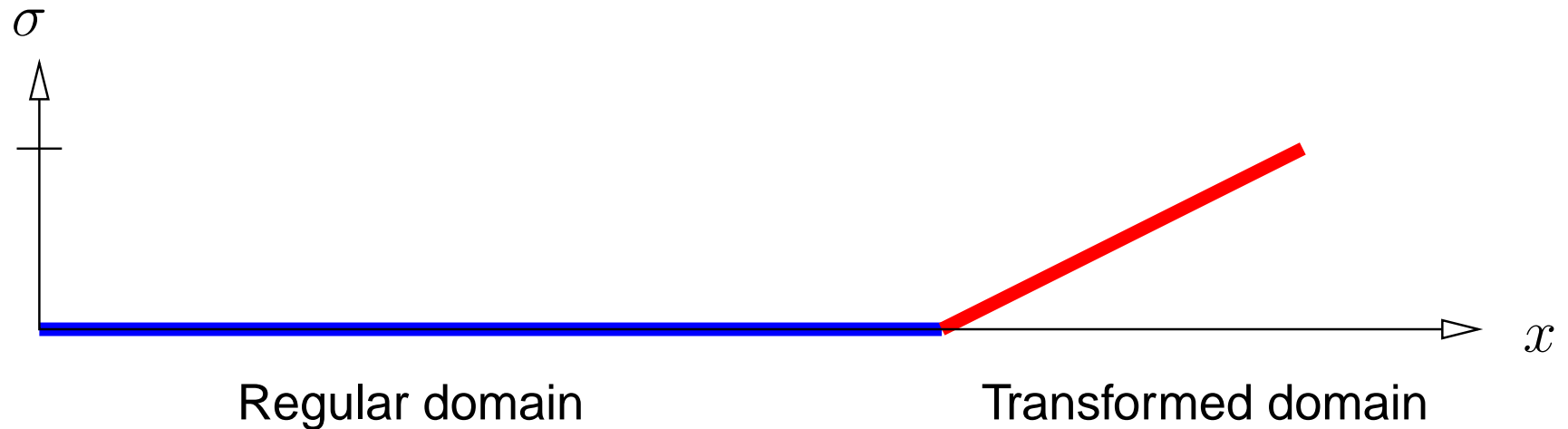


$$\frac{d\tilde{x}}{dx} = \lambda(x) \text{ where } \lambda(s) = 1 - i\sigma(s)$$

$$\frac{1}{\lambda} \frac{d}{dx} \left(\frac{1}{\lambda} \frac{d\hat{u}}{dx} \right) + k^2 \hat{u} = 0$$

$$\hat{u} = c_{\text{out}} \exp \left(-k \int_0^x \sigma(s) ds \right) e^{-ikx} + c_{\text{in}} \exp \left(k \int_0^x \sigma(s) ds \right) e^{ikx}$$

1-D model problem with PML

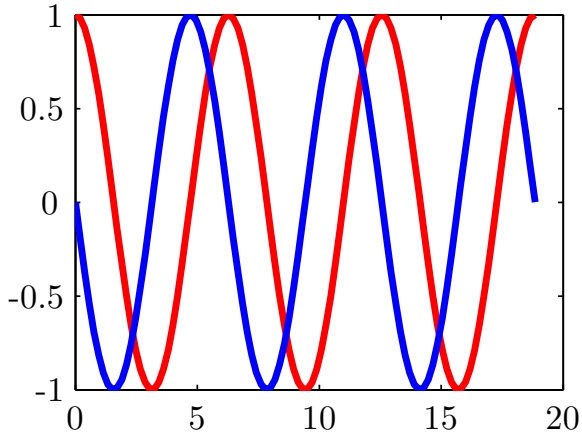


If solution clamped at $x = L$ then

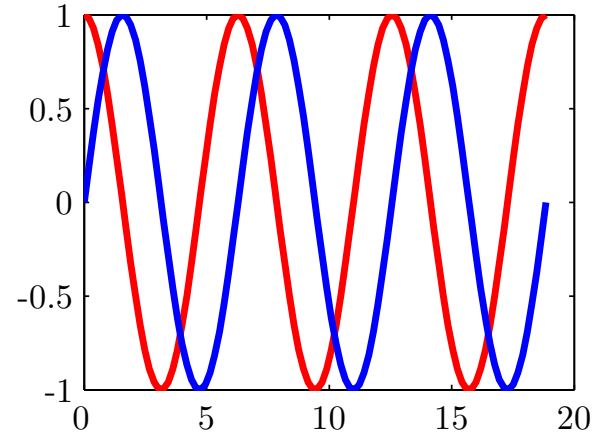
$$\frac{c_{\text{in}}}{c_{\text{out}}} = O(e^{-k\gamma}) \text{ where } \gamma = \int_0^L \sigma(s) ds$$

1-D model problem illustrated

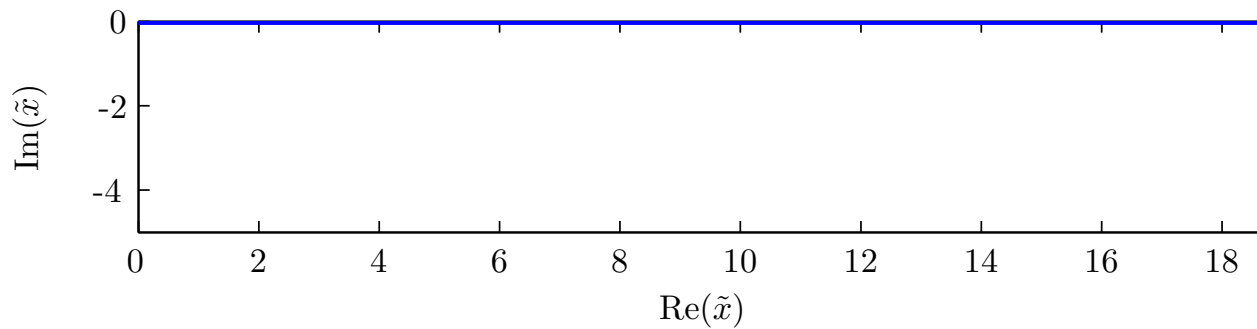
Outgoing $\exp(-i\tilde{x})$



Incoming $\exp(i\tilde{x})$

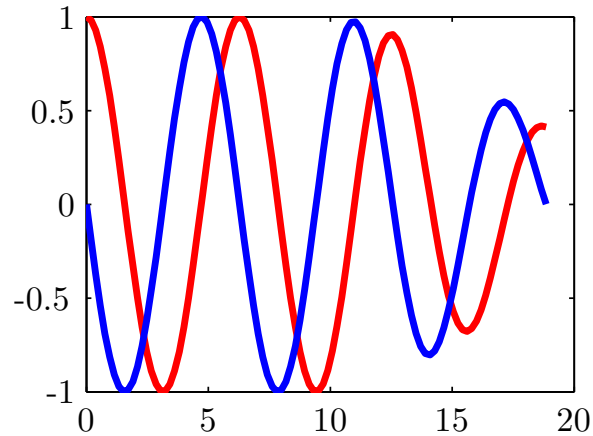


Transformed coordinate

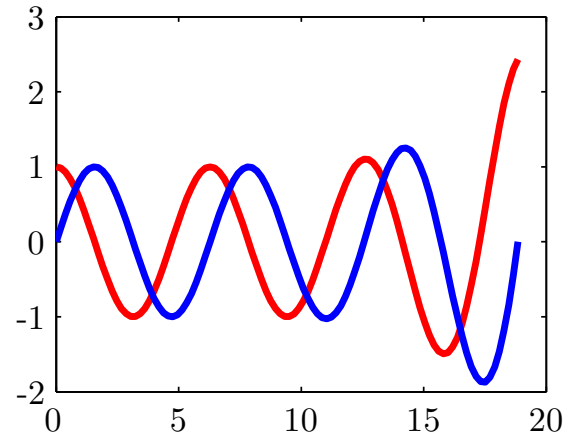


1-D model problem illustrated

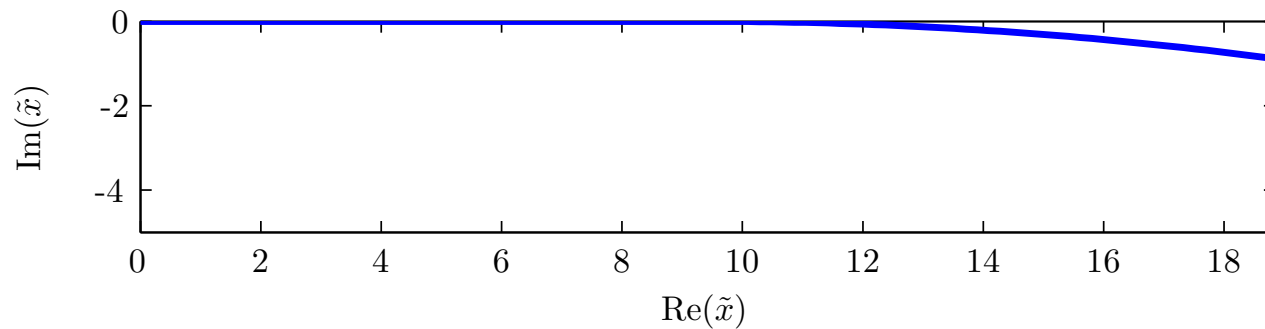
Outgoing $\exp(-i\tilde{x})$



Incoming $\exp(i\tilde{x})$

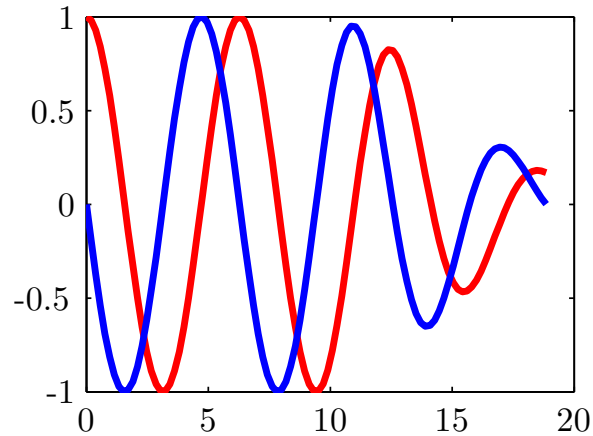


Transformed coordinate

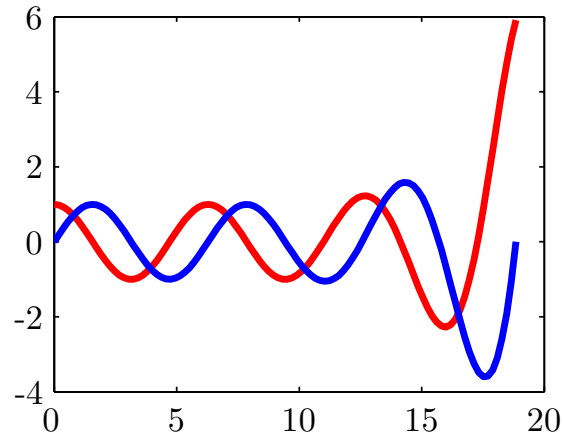


1-D model problem illustrated

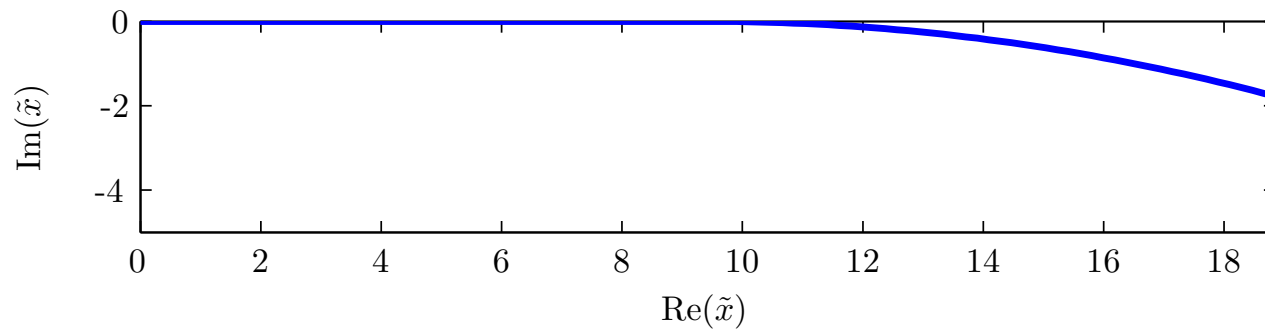
Outgoing $\exp(-i\tilde{x})$



Incoming $\exp(i\tilde{x})$

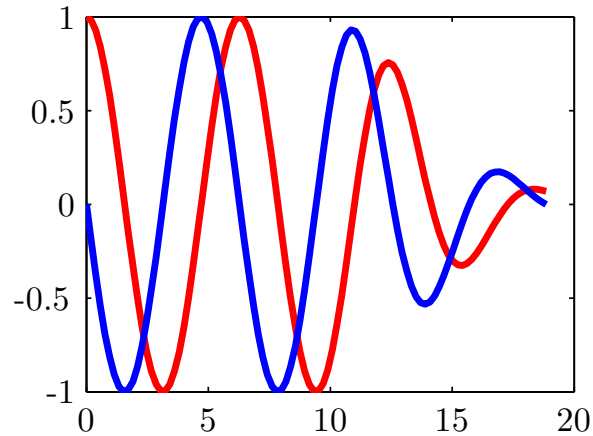


Transformed coordinate

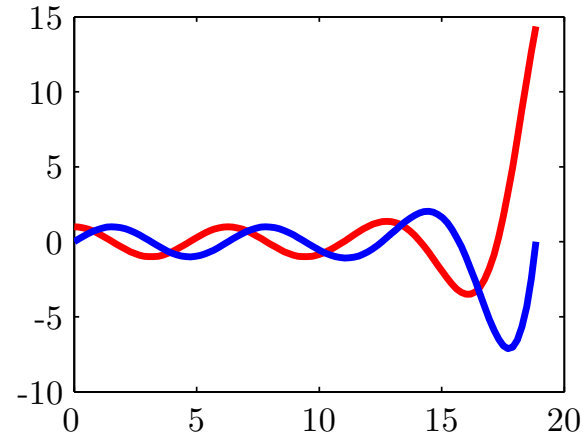


1-D model problem illustrated

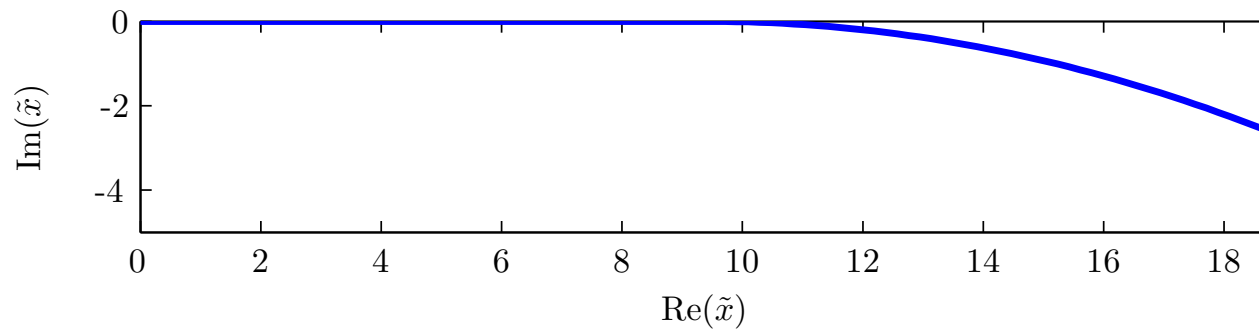
Outgoing $\exp(-i\tilde{x})$



Incoming $\exp(i\tilde{x})$

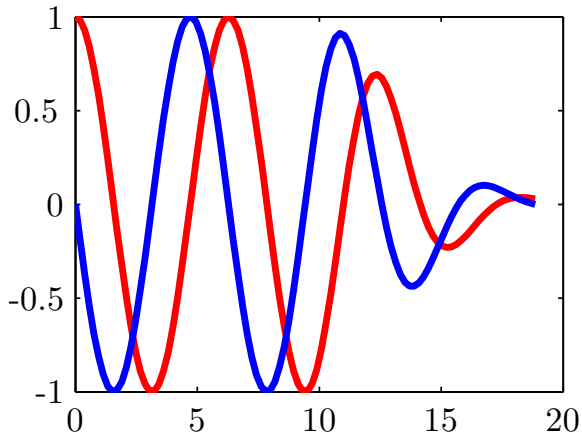


Transformed coordinate

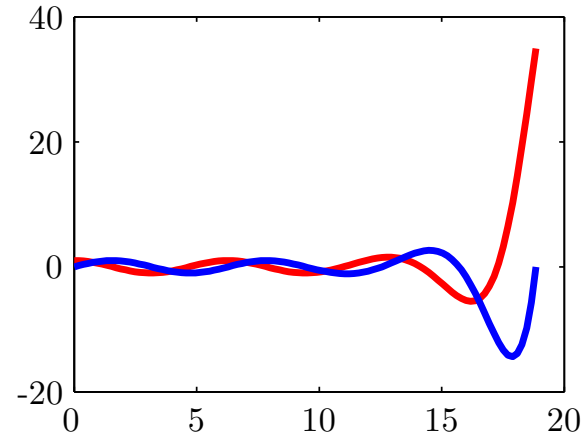


1-D model problem illustrated

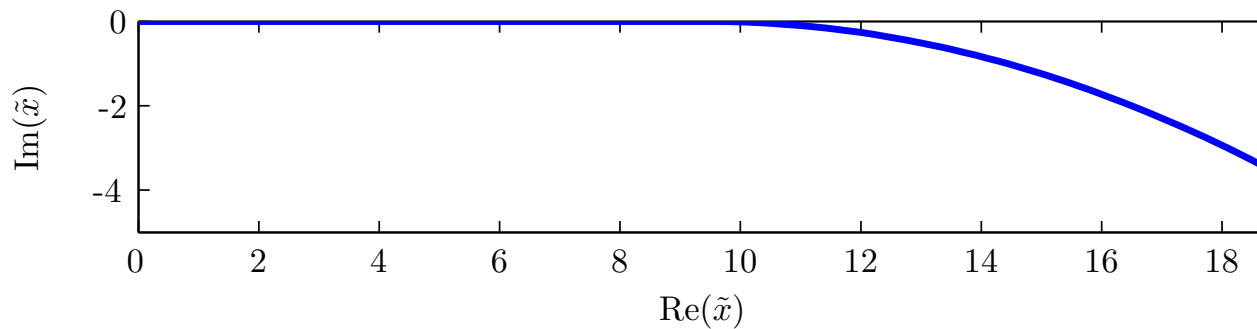
Outgoing $\exp(-i\tilde{x})$



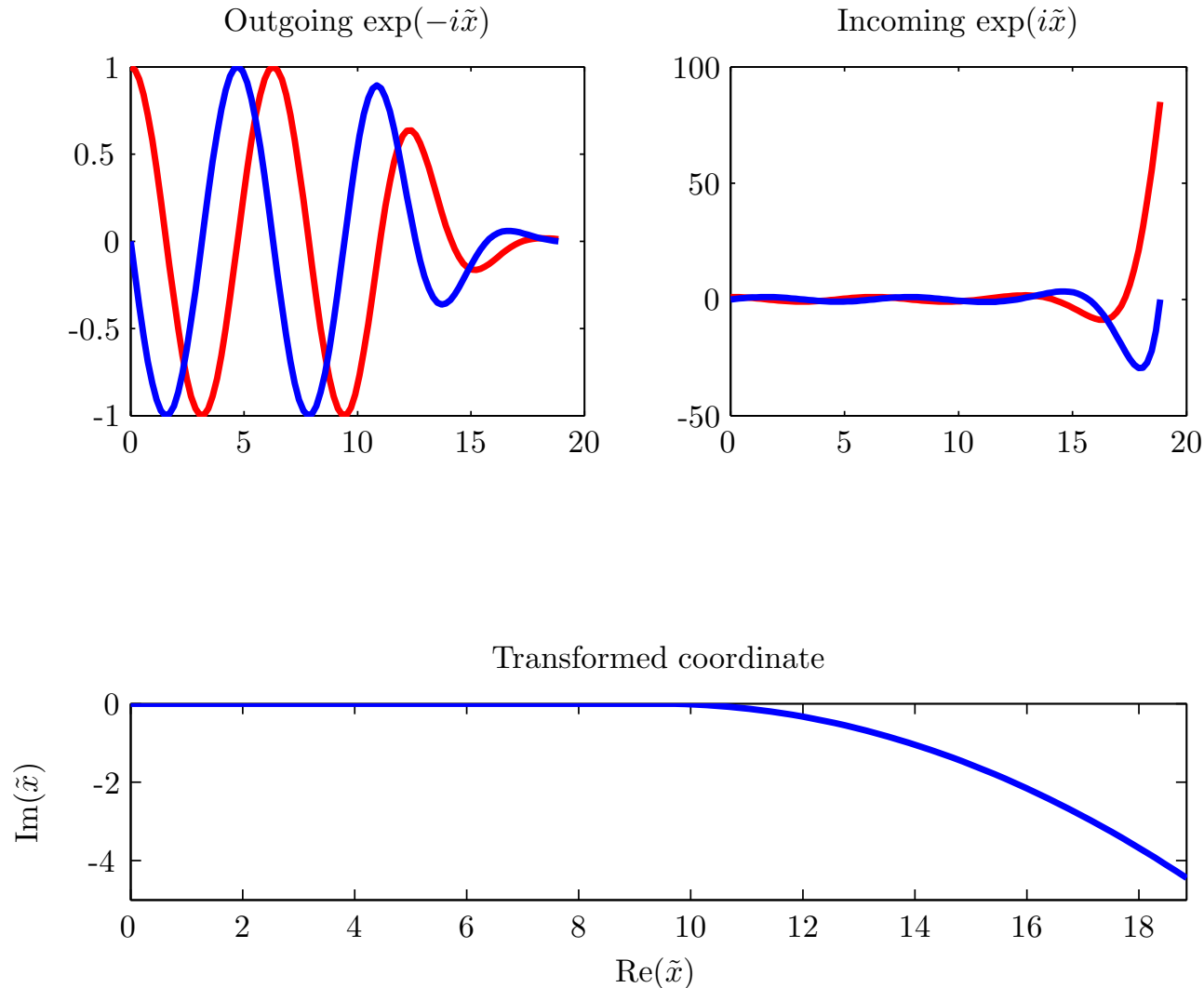
Incoming $\exp(i\tilde{x})$



Transformed coordinate



1-D model problem illustrated



Clamp solution at transformed end to isolate outgoing wave.

Elastic PMLs

$$\int_{\Omega} \epsilon(w) : \mathbb{C} : \epsilon(u) d\Omega - \omega^2 \int_{\Omega} \rho w \cdot u d\Omega = \int_{\Gamma} w \cdot t_n d\Gamma$$

$$\epsilon(u) = \left(\frac{\partial u}{\partial x} \right)^s$$

- Start from standard weak form

Elastic PMLs

$$\int_{\tilde{\Omega}} \tilde{\epsilon}(w) : \mathbf{C} : \tilde{\epsilon}(u) d\tilde{\Omega} - \omega^2 \int_{\tilde{\Omega}} \rho w \cdot u d\tilde{\Omega} = \int_{\Gamma} w \cdot t_n d\Gamma$$

$$\tilde{\epsilon}(u) = \left(\frac{\partial u}{\partial \tilde{x}} \right)^s = \left(\frac{\partial u}{\partial x} \Lambda^{-1} \right)^s$$

- Start from standard weak form
- Introduce transformed \tilde{x} with $\frac{\partial \tilde{x}}{\partial x} = \Lambda$

Elastic PMLs

$$\int_{\Omega} \tilde{\epsilon}(w) : \mathbf{C} : \tilde{\epsilon}(u) J_{\Lambda} d\Omega - \omega^2 \int_{\Omega} \rho w \cdot u J_{\Lambda} d\Omega = \int_{\Gamma} w \cdot t_n d\Gamma$$

$$\tilde{\epsilon}(u) = \left(\frac{\partial u}{\partial \tilde{x}} \right)^s = \left(\frac{\partial u}{\partial x} \Lambda^{-1} \right)^s$$

- Start from standard weak form
- Introduce transformed \tilde{x} with $\frac{\partial \tilde{x}}{\partial x} = \Lambda$
- Map back to reference system ($J_{\Lambda} = \det(\Lambda)$)

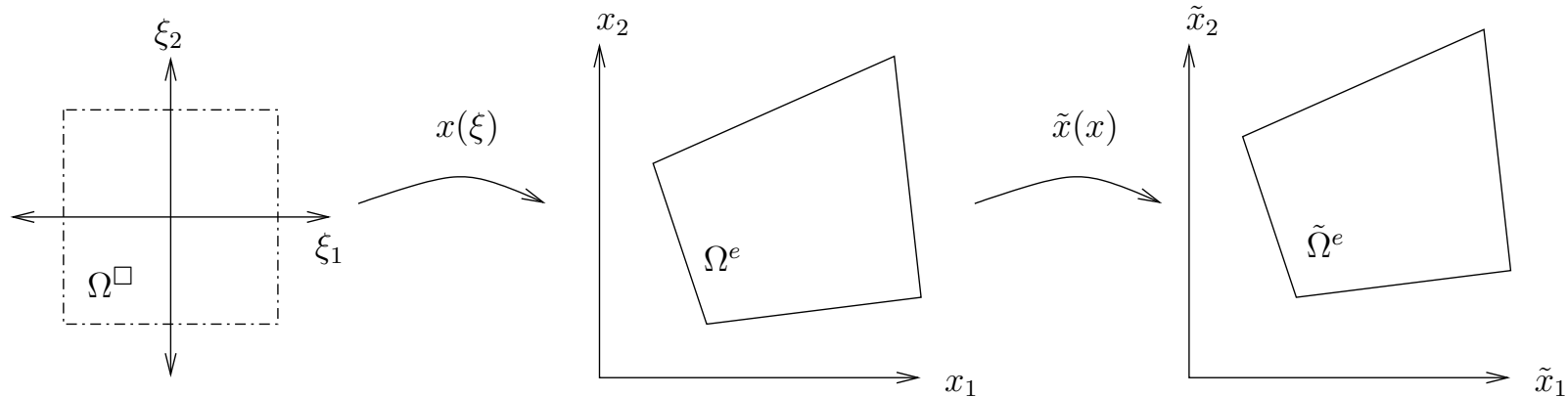
Elastic PMLs

$$\int_{\Omega} \tilde{\epsilon}(w) : \mathbf{C} : \tilde{\epsilon}(u) J_{\Lambda} d\Omega - \omega^2 \int_{\Omega} \rho w \cdot u J_{\Lambda} d\Omega = \int_{\Gamma} w \cdot t_n d\Gamma$$

$$\tilde{\epsilon}(u) = \left(\frac{\partial u}{\partial \tilde{x}} \right)^s = \left(\frac{\partial u}{\partial x} \Lambda^{-1} \right)^s$$

- Start from standard weak form
- Introduce transformed \tilde{x} with $\frac{\partial \tilde{x}}{\partial x} = \Lambda$
- Map back to reference system ($J_{\Lambda} = \det(\Lambda)$)
- All terms are symmetric in w and u

Finite element implementation



- Combine PML and isoparametric mappings

$$\mathbf{k}^e = \int_{\Omega^\square} \tilde{\mathbf{B}}^T \mathbf{D} \tilde{\mathbf{B}} \tilde{J} d\Omega^\square$$

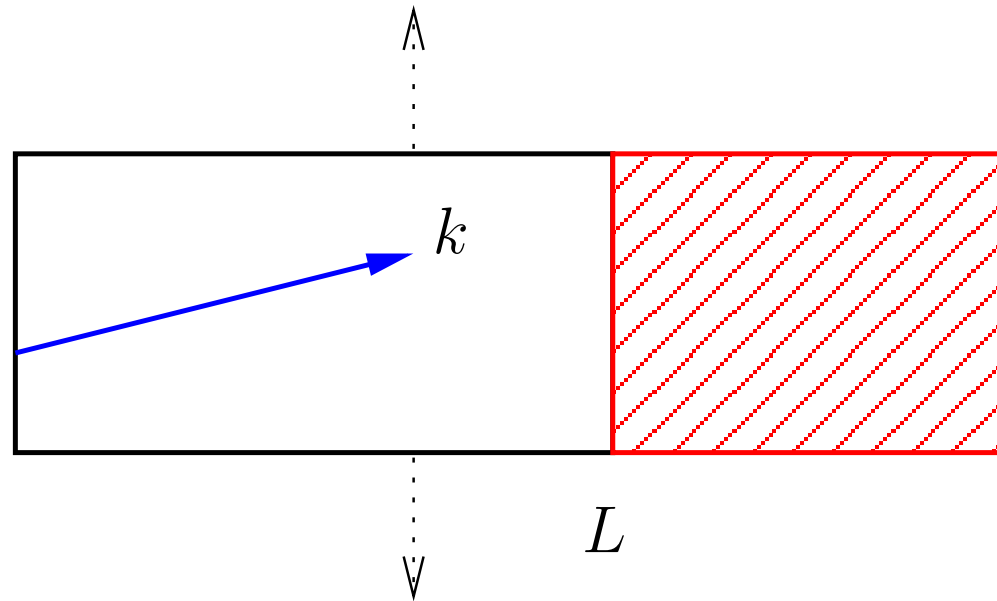
$$\mathbf{m}^e = \left(\int_{\Omega^\square} \rho \mathbf{N}^T \mathbf{N} \tilde{J} d\Omega^\square \right)$$

- Matrices are *complex symmetric*

Outline

- Electromechanical resonators and RF MEMS
- Damping and quality of resonance
- Anchor losses and Perfectly Matched Layers
- Analysis of the discretized PMLs
 - A two-dimensional model problem
 - Analysis of discrete reflection
 - Choice of PML parameters
- Complex symmetry and structured model reduction
- Analysis of a disk resonator
- Conclusions

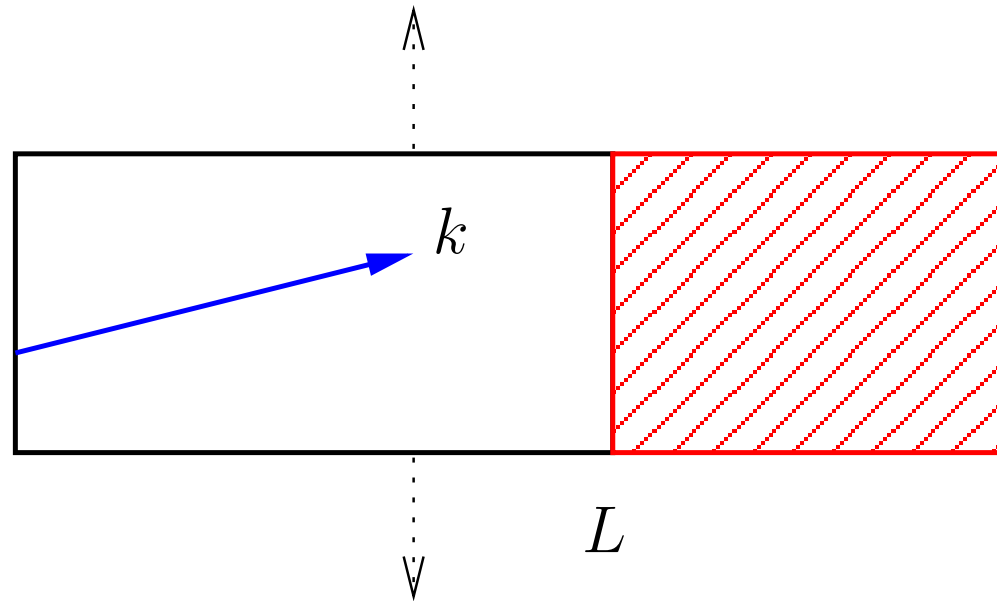
Continuum 2D model problem



$$\lambda(x) = \begin{cases} 1 - i\beta|x - L|^p, & x > L \\ 1 & x \leq L. \end{cases}$$

$$\frac{1}{\lambda} \frac{\partial}{\partial x} \left(\frac{1}{\lambda} \frac{\partial u}{\partial x} \right) + \frac{\partial^2 u}{\partial y^2} + k^2 u = 0$$

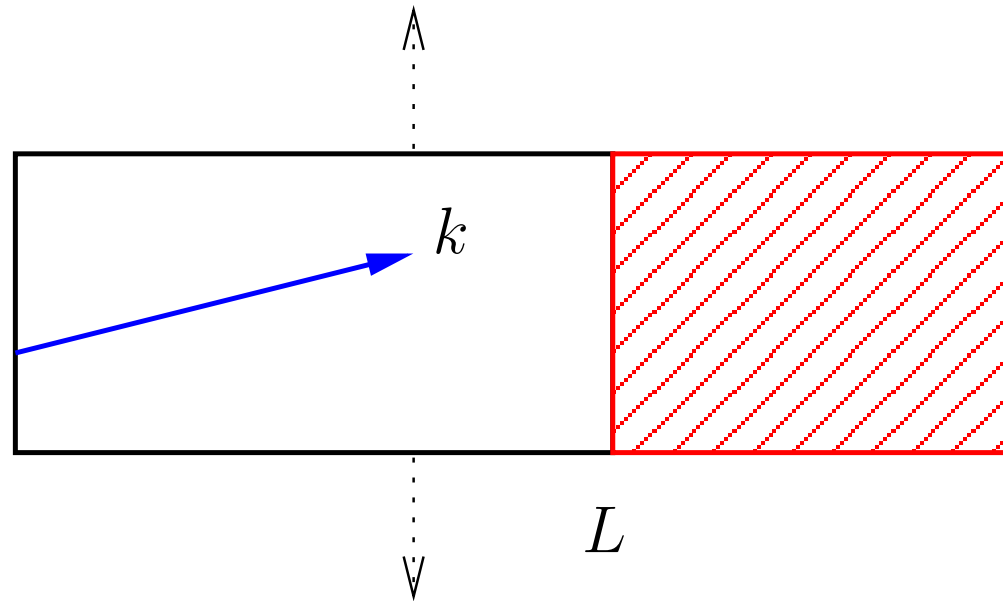
Continuum 2D model problem



$$\lambda(x) = \begin{cases} 1 - i\beta|x - L|^p, & x > L \\ 1 & x \leq L. \end{cases}$$

$$\frac{1}{\lambda} \frac{\partial}{\partial x} \left(\frac{1}{\lambda} \frac{\partial u}{\partial x} \right) - k_y^2 u + k^2 u = 0$$

Continuum 2D model problem

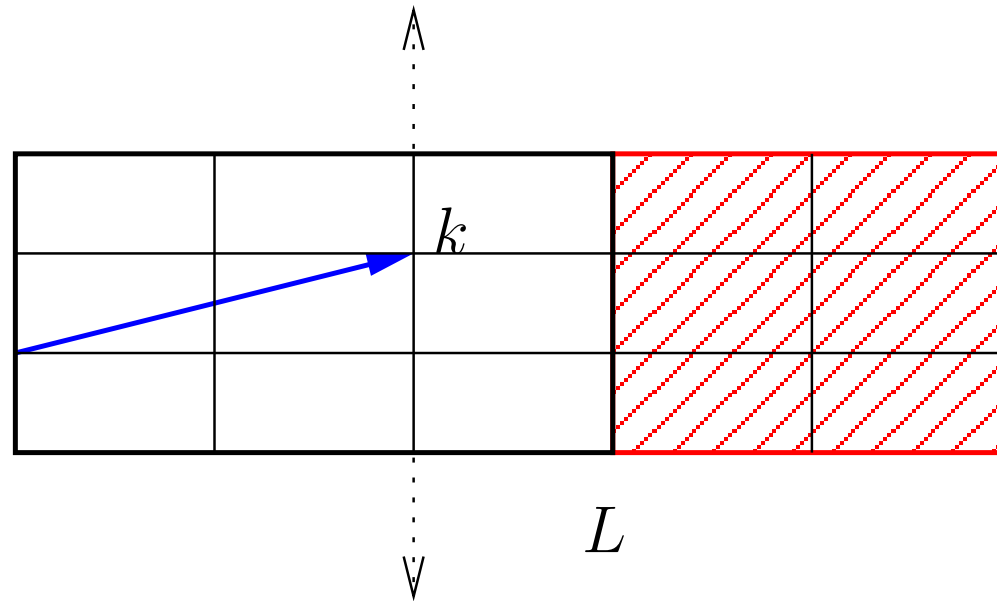


$$\lambda(x) = \begin{cases} 1 - i\beta|x - L|^p, & x > L \\ 1 & x \leq L. \end{cases}$$

$$\frac{1}{\lambda} \frac{\partial}{\partial x} \left(\frac{1}{\lambda} \frac{\partial u}{\partial x} \right) + k_x^2 u = 0$$

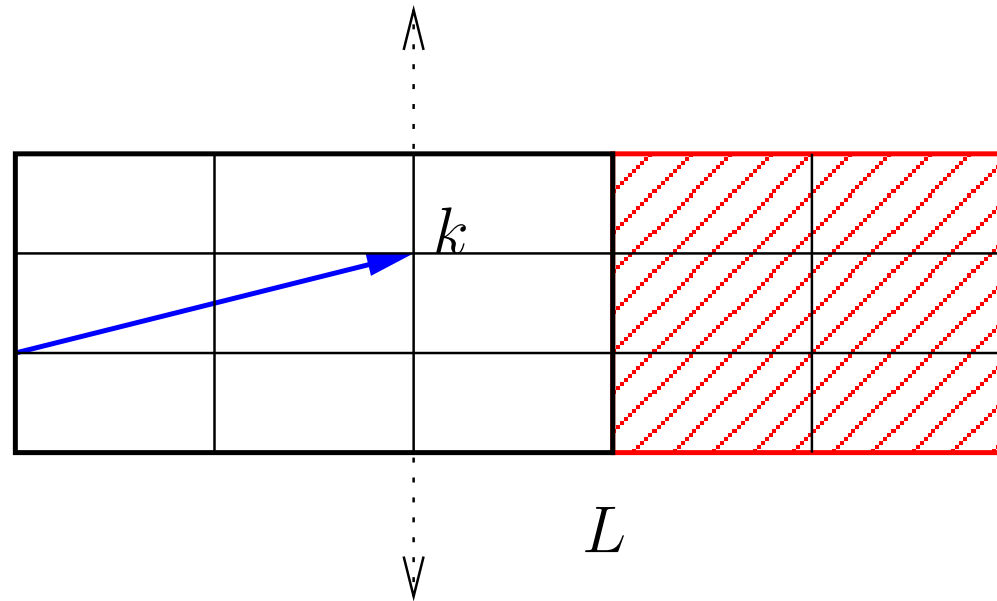
1D problem, reflection of $O(e^{-k_x \gamma})$

Discrete 2D model problem



- Discrete Fourier transform in y
- Solve numerically in x
- Project solution onto infinite space traveling modes
- Extension of Collino and Monk (1998)

Nondimensionalization



$$\lambda(x) = \begin{cases} 1 - i\beta|x - L|^p, & x > L \\ 1 & x \leq L. \end{cases}$$

Rate of stretching:

$$\beta h^p$$

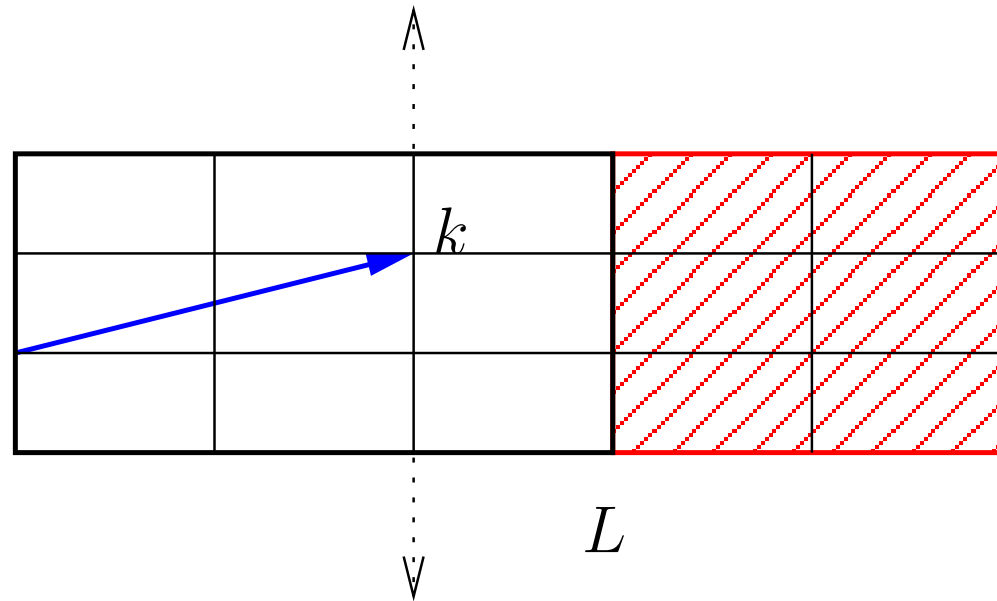
Elements per wave:

$$(k_x h)^{-1} \text{ and } (k_y h)^{-1}$$

Elements through the PML:

$$N$$

Nondimensionalization



$$\lambda(x) = \begin{cases} 1 - i\beta|x - L|^p, & x > L \\ 1 & x \leq L. \end{cases}$$

Rate of stretching:

$$\beta h^p$$

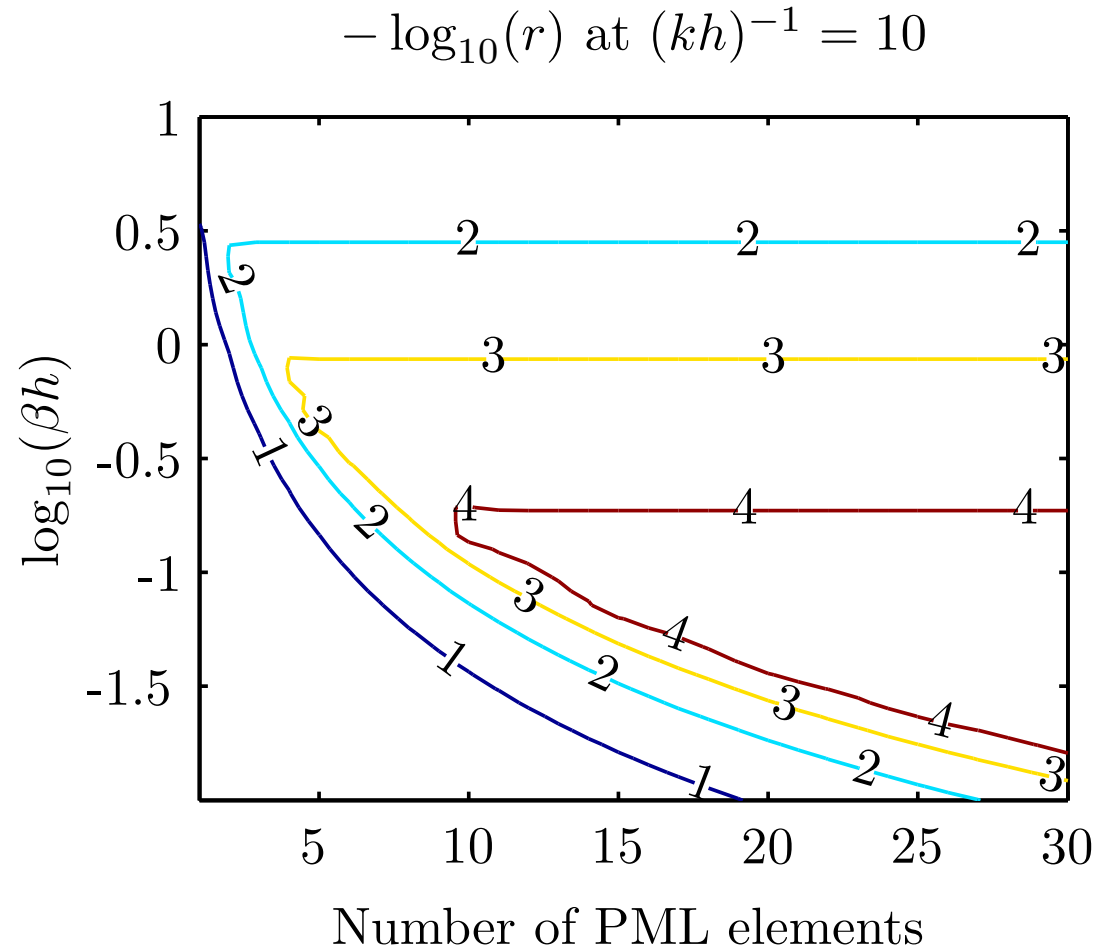
Elements per wave:

$$(k_x h)^{-1} \quad \text{and} \quad (k_y h)^{-1}$$

Elements through the PML:

$$N$$

Discrete reflection behavior



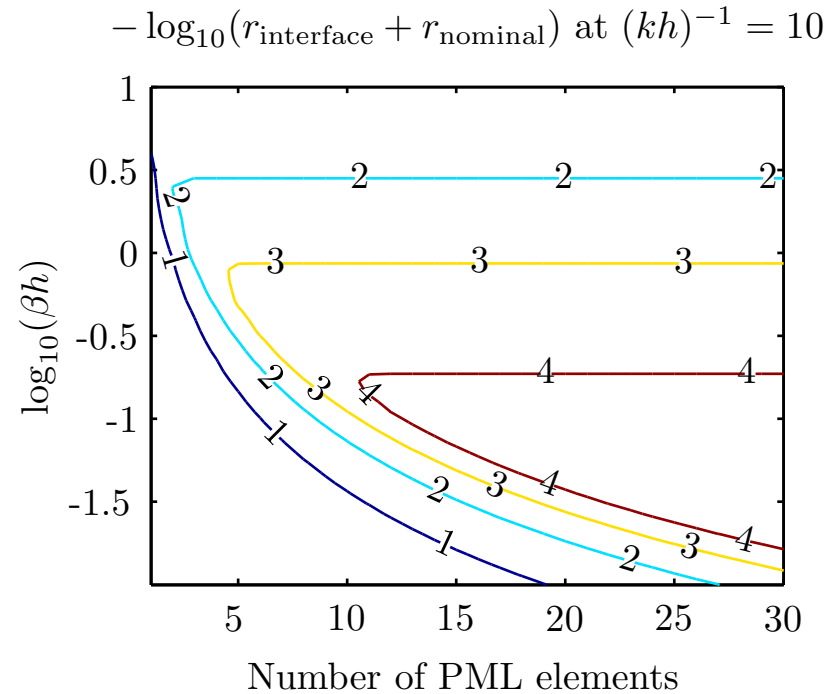
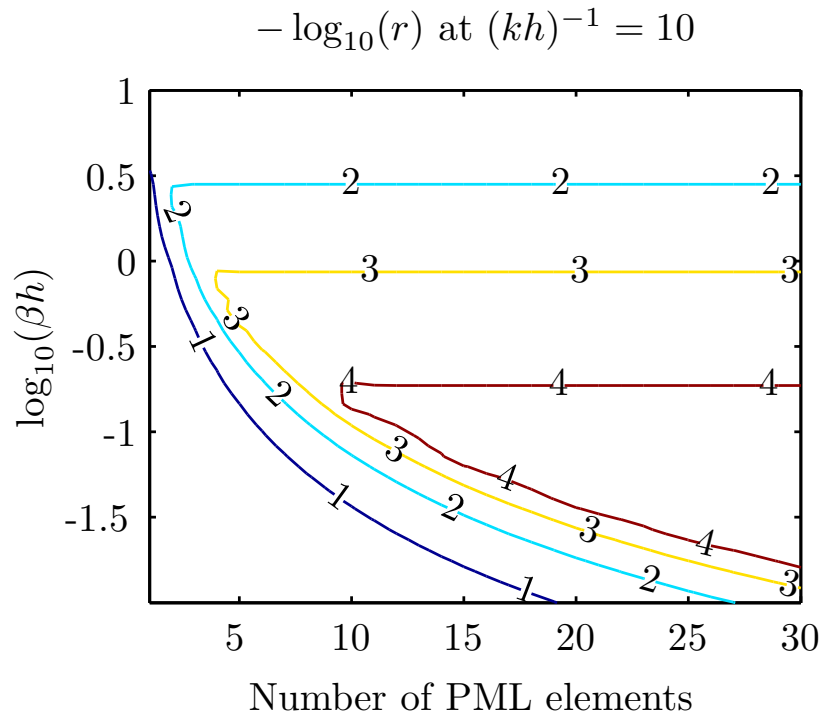
Quadratic elements, $p = 1$, $(k_x h)^{-1} = 10$

Discrete reflection decomposition

Model discrete reflection as two parts:

- Far-end reflection (clamping reflection)
 - Approximated well by continuum calculation
 - Grows as $(k_x h)^{-1}$ grows
- Interface reflection
 - Discrete effect: mesh does not resolve decay
 - Does not depend on N
 - Grows as $(k_x h)^{-1}$ shrinks

Discrete reflection behavior



Quadratic elements, $p = 1$, $(k_x h)^{-1} = 10$

- Model does well at predicting actual reflection
- Similar picture for other wavelengths, element types, stretch functions

Choosing PML parameters

- Discrete reflection dominated by
 - Interface reflection when k_x large
 - Far-end reflection when k_x small
- Heuristic for PML parameter choice
 - Choose an acceptable reflection level
 - Choose β based on interface reflection at k_x^{\max}
 - Choose length based on far-end reflection at k_x^{\min}

Outline

- Electromechanical resonators and RF MEMS
- Damping and quality of resonance
- Anchor losses and Perfectly Matched Layers
- Analysis of the discretized PMLs
- Complex symmetry and structured model reduction
 - Krylov subspace projections
 - Structure-preserving eigencomputations
 - Structure-preserving model reduction
- Analysis of a disk resonator
- Conclusions

Eigenvalues and model reduction

Want to know about the transfer function $H(\omega)$:

$$H(\omega) = p^T (K - \omega^2 M)^{-1} b$$

Can either

- Locate poles of H (eigenvalues of (K, M))
 - Determine $Q = \frac{|\omega|}{2 \operatorname{Im}(\omega)}$
- Plot H in a frequency range (Bode plot)

Solve both problems with the same tool:

Krylov subspace projections

Projecting via Arnoldi

Build a Krylov subspace basis by shift-invert Arnoldi

- Choose shift σ in frequency range of interest
- Form and factor $K_{\text{shift}} = K - \sigma^2 M$
- Use Arnoldi to build an orthonormal basis V for

$$\mathcal{K}_n = \text{span}\{u_0, K_{\text{shift}}^{-1}u_0, \dots, K_{\text{shift}}^{-(n-1)}u_0\}$$

Compute eigenvalues and reduced models from projection

- Compute eigenvalues from (V^*KV, V^*MV)
- Approximate $H(\omega)$ by Galerkin projection

$$H(\omega) \approx (V^*p)^*(V^*KV - \omega^2V^*MV)^{-1}(V^*b)$$

Accurate eigenvalues

- *Hermitian* systems: Rayleigh-Ritz is optimal
 - Raleigh quotient is stationary at eigenvectors

$$\rho(v) = \frac{v^* K v}{v^* M v}$$

- First-order accurate eigenvectors \implies second-order accurate eigenvalues
- Can we obtain optimal accuracy for PML eigenvalues?

Accurate eigenvalues

- PML matrices are *complex symmetric*
 - Modified RQ is stationary at eigenvectors

$$\theta(v) = \frac{v^T K v}{v^T M v}$$

- \implies second-order accurate eigenvalues
- Hochstenbach and Arbenz, 2004

Accurate model reduction

- Accurate eigenvalues from v and \bar{v} together
- Accurate model reduction in the same way
 - Build new projection basis from V :

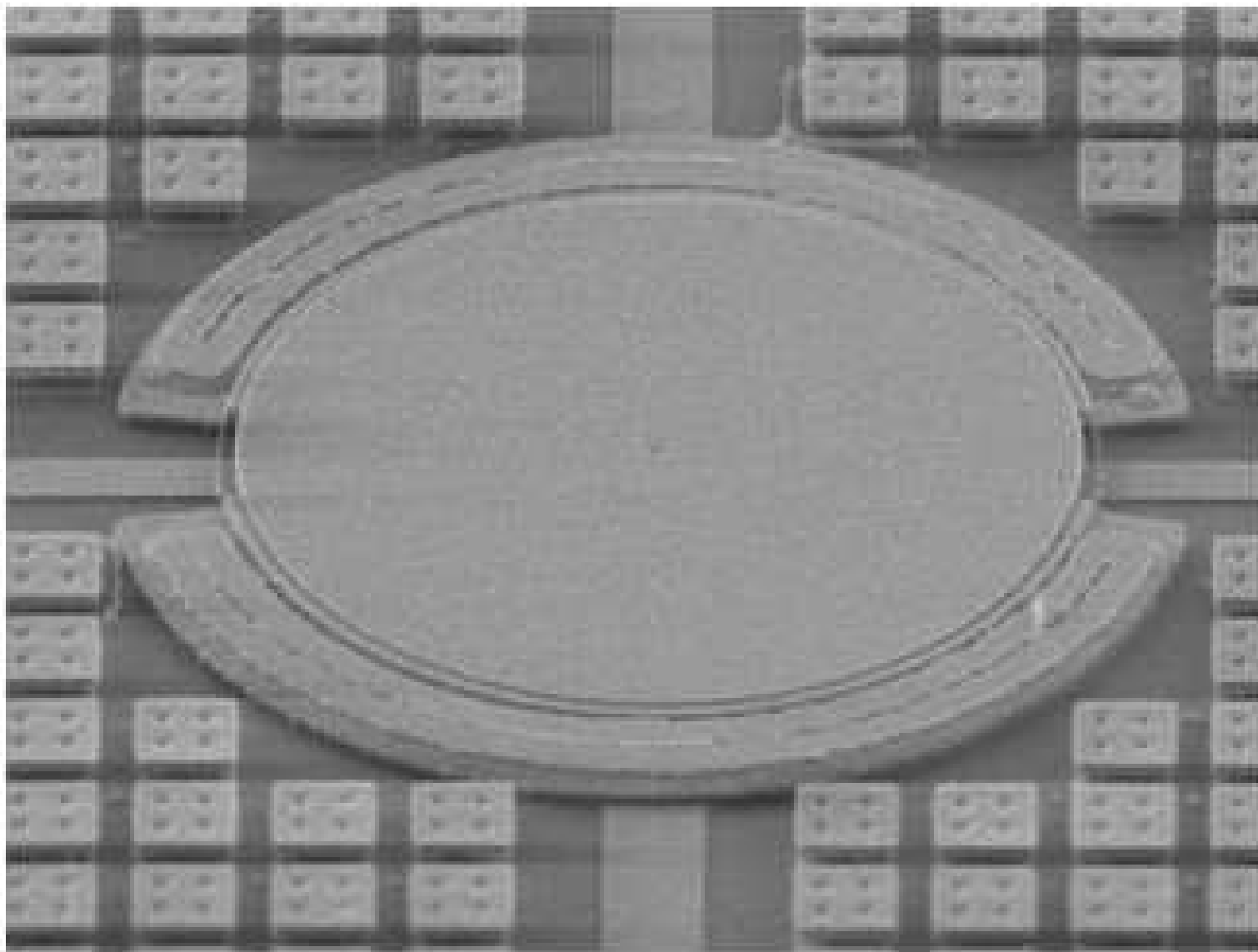
$$W = \text{orth}[\text{Re}(V), \text{Im}(V)]$$

- $\text{span}(W)$ contains both \mathcal{K}_n and $\bar{\mathcal{K}}_n$
 - Double convergence vs projection with V
- W is a real-valued basis
 - Projected system remains complex symmetric

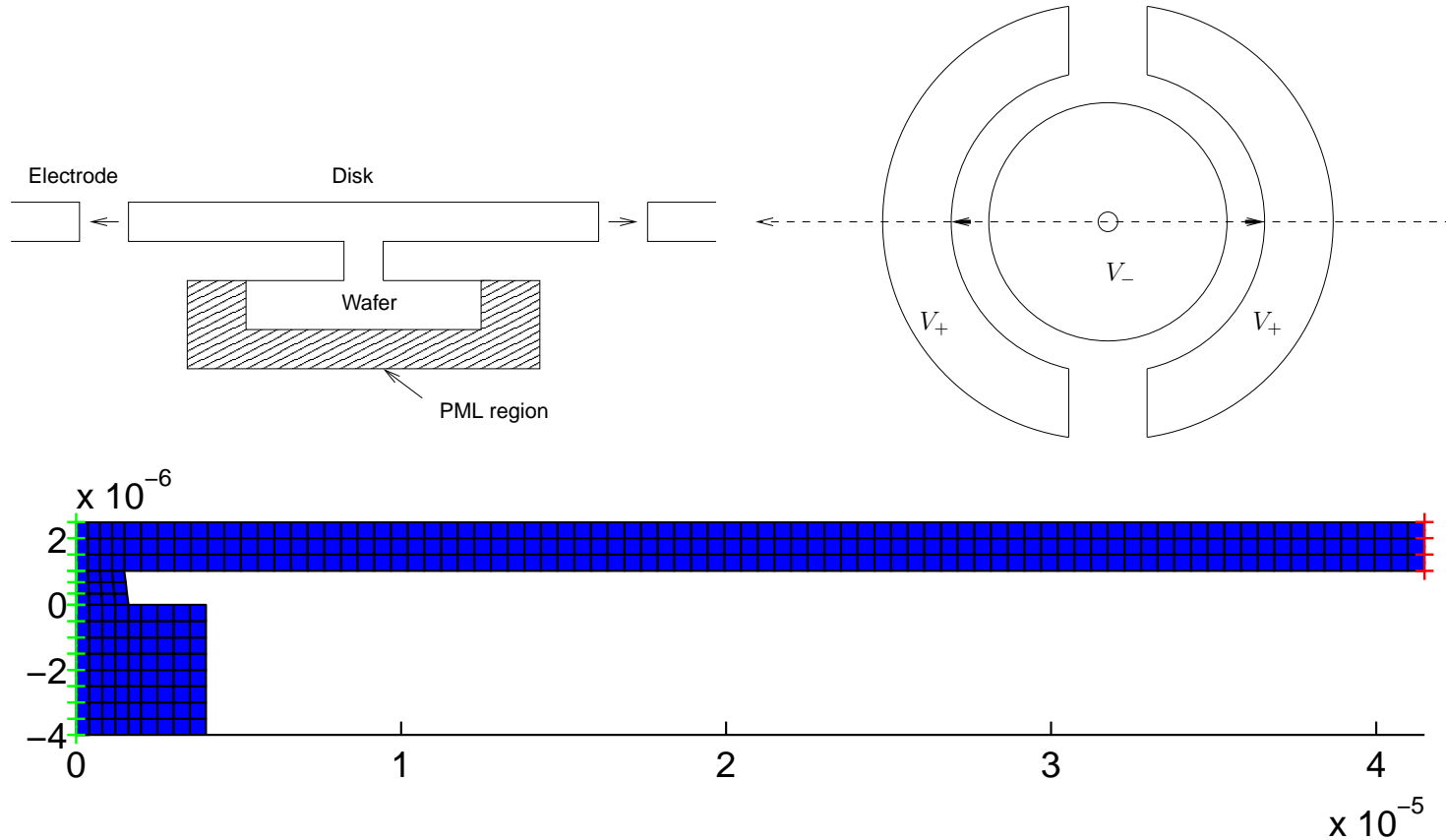
Outline

- Electromechanical resonators and RF MEMS
- Damping and quality of resonance
- Anchor losses and Perfectly Matched Layers
- Analysis of the discretized PMLs
- Complex symmetry and structured model reduction
- Analysis of a disk resonator
 - Accuracy of the numerics
 - Description of the loss mechanism
 - Sensitivity to fabrication variations
- Conclusions

Disk resonator simulations

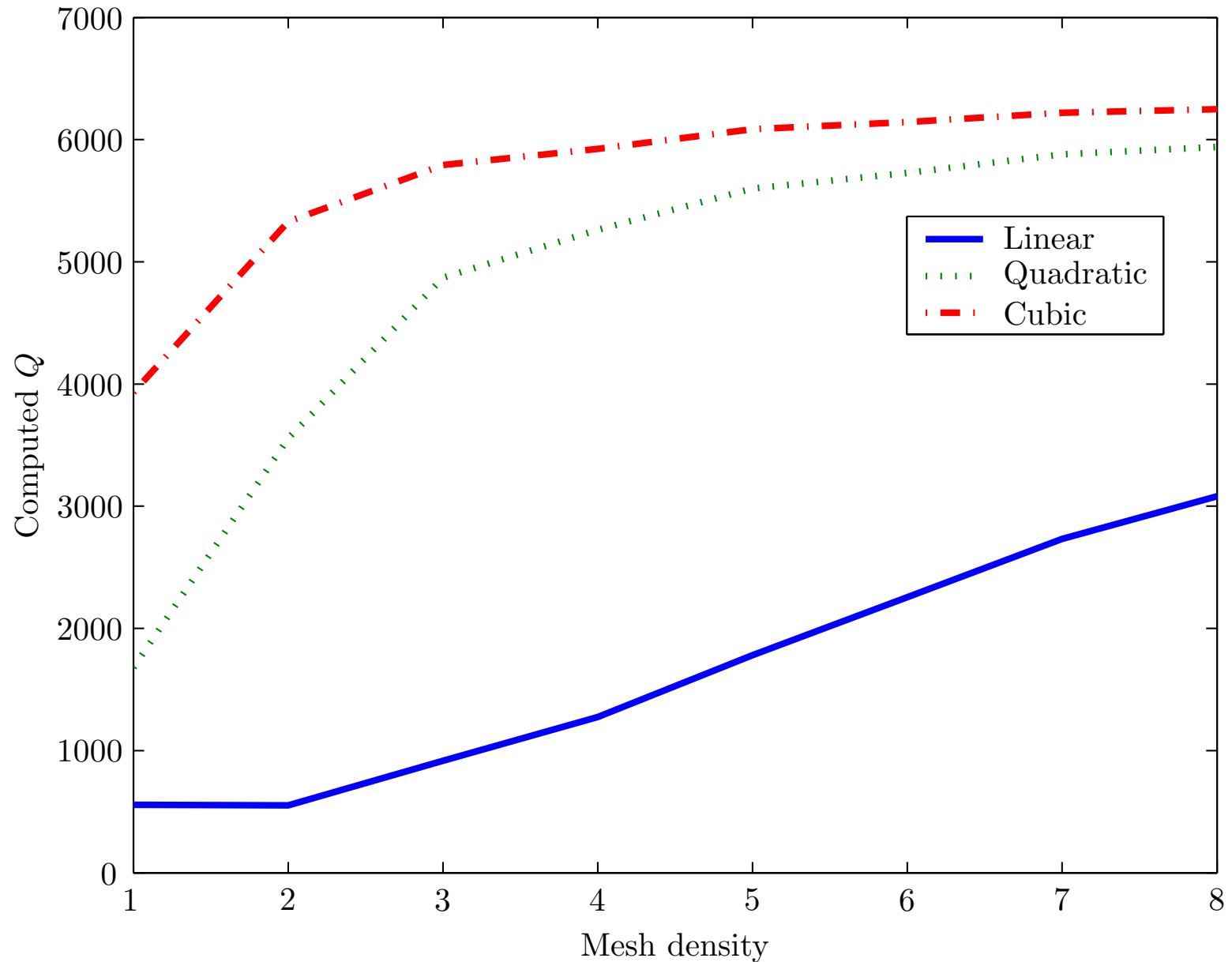


Disk resonator mesh

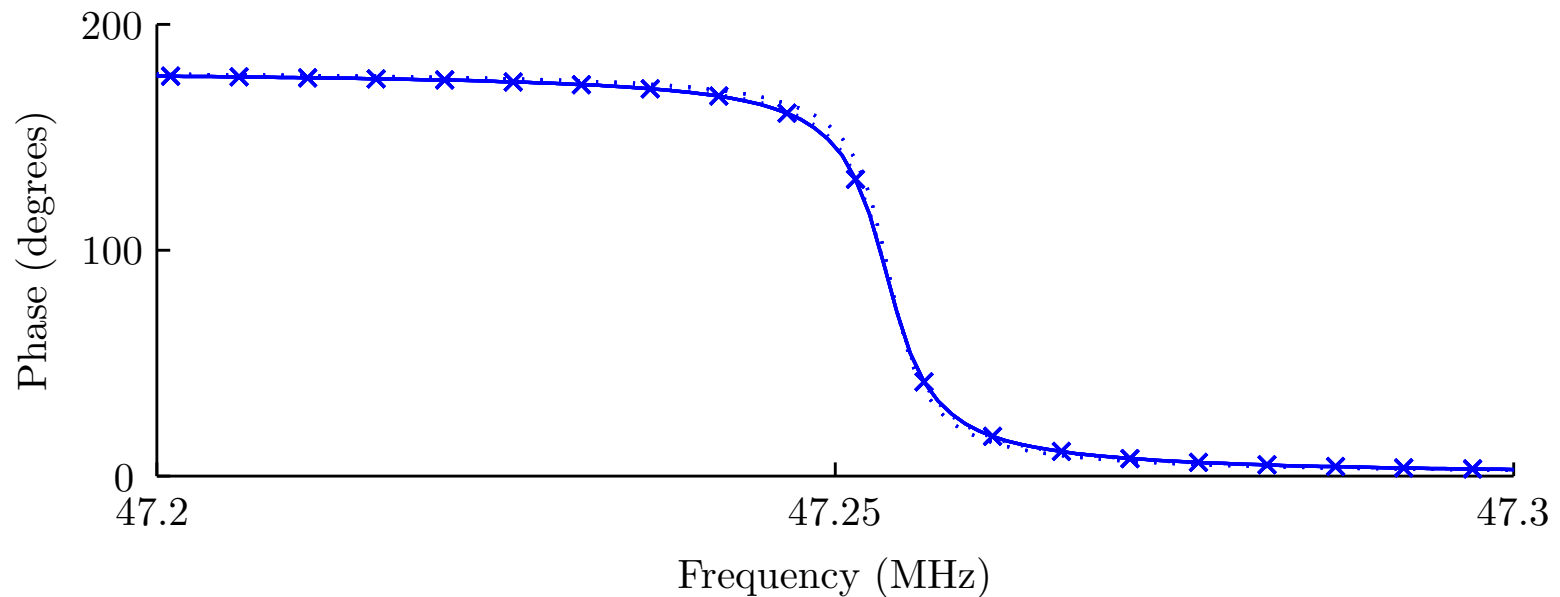
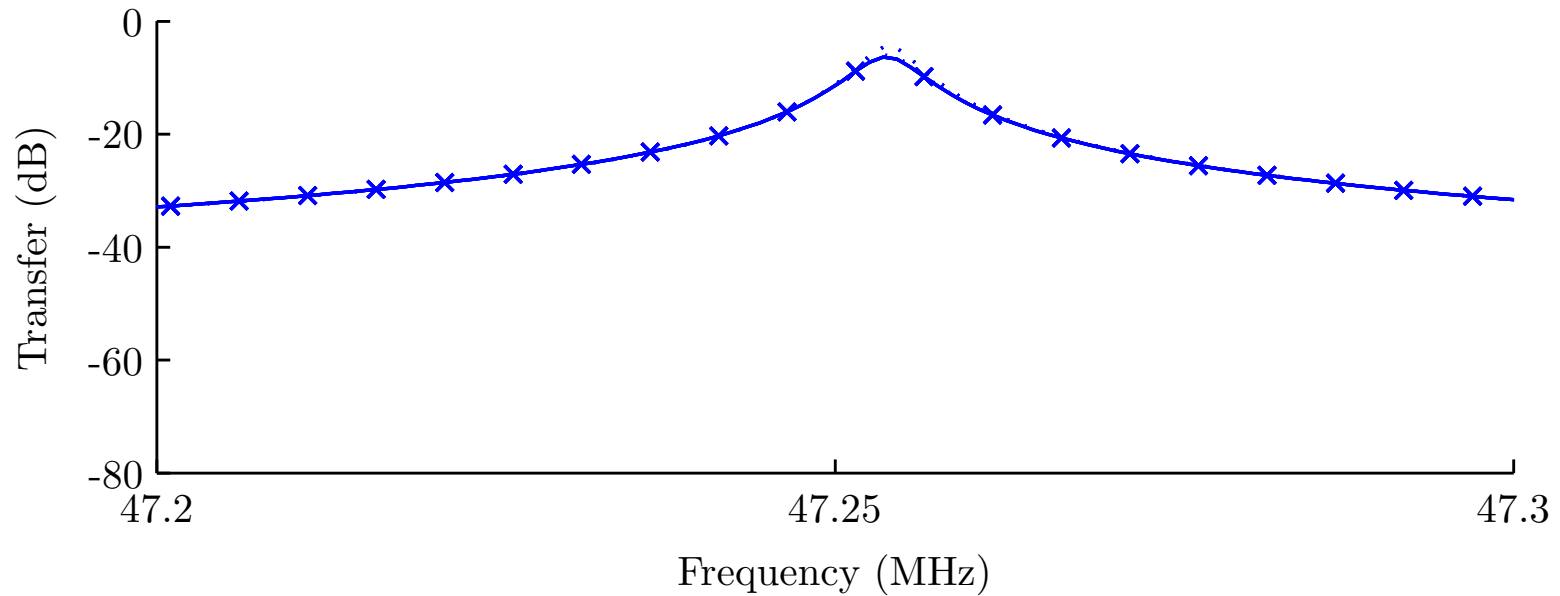


- Axisymmetric model with bicubic mesh
- About 10K nodal points in converged calculation

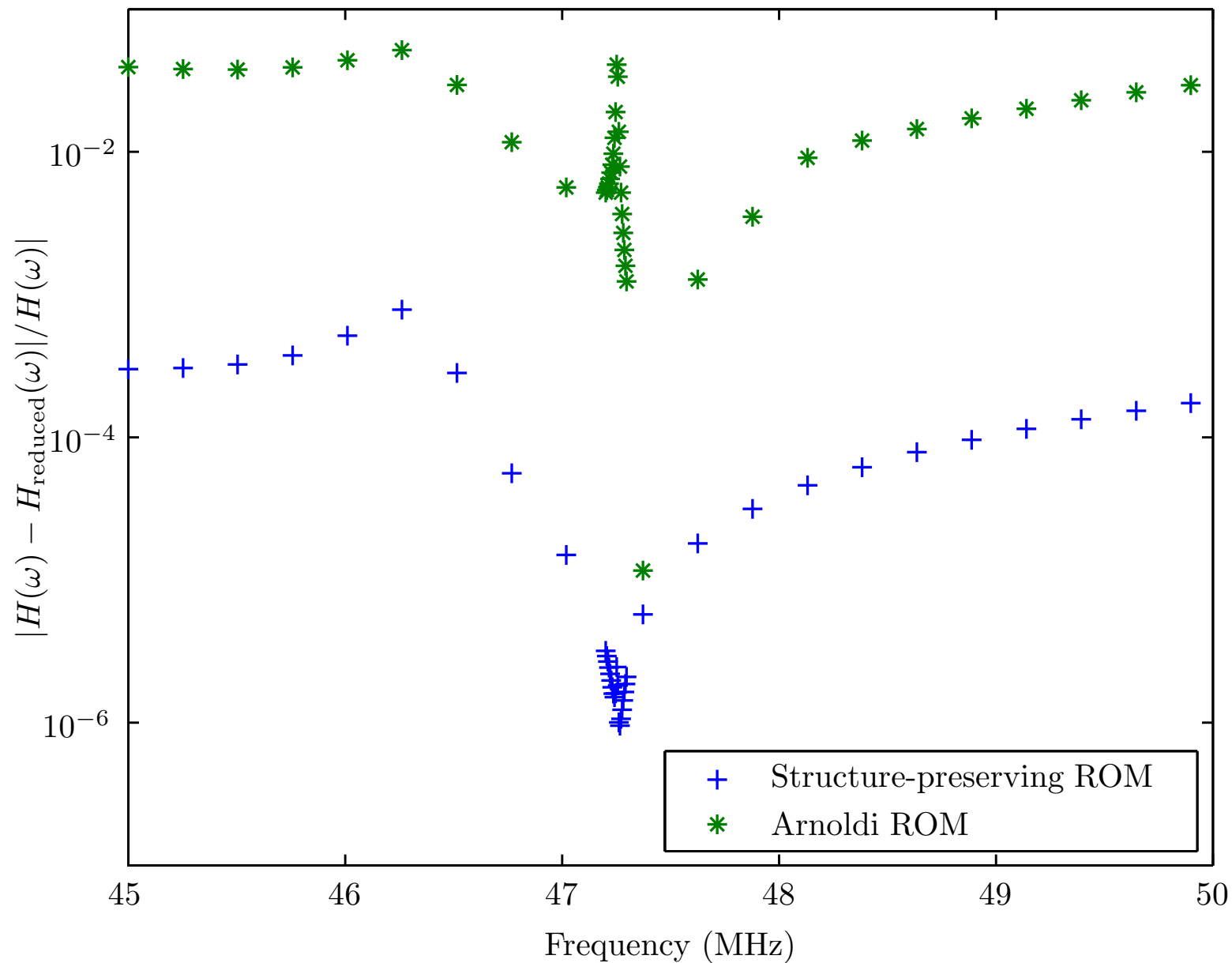
Mesh convergence



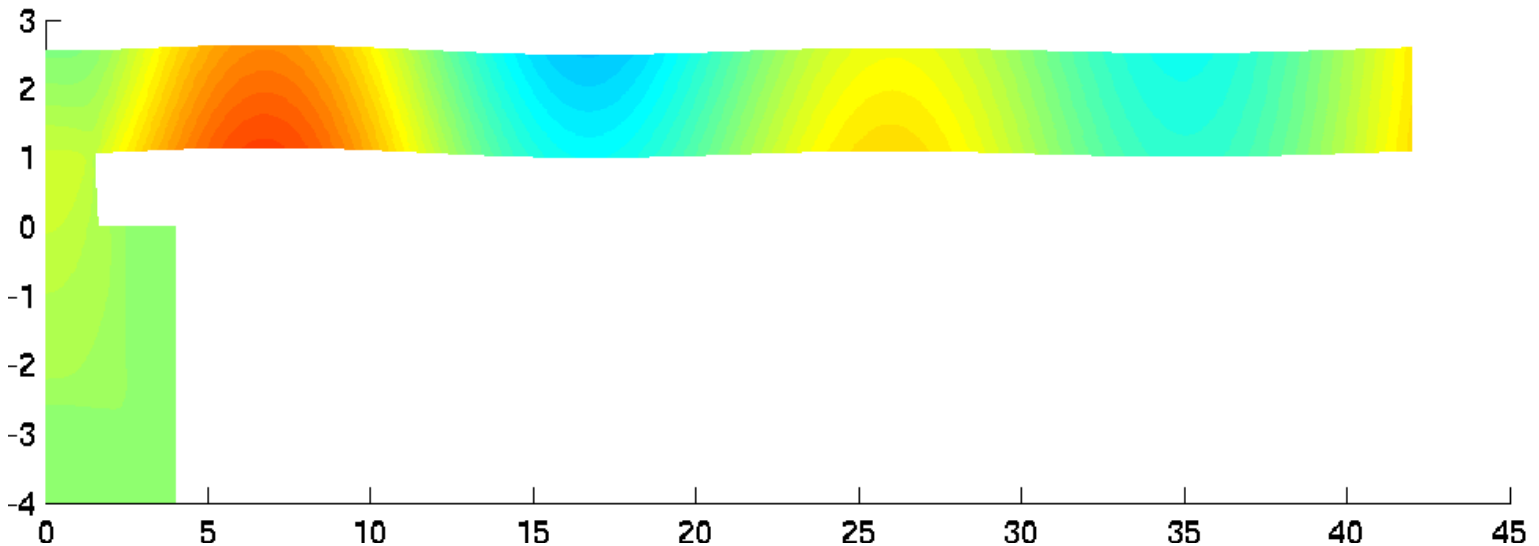
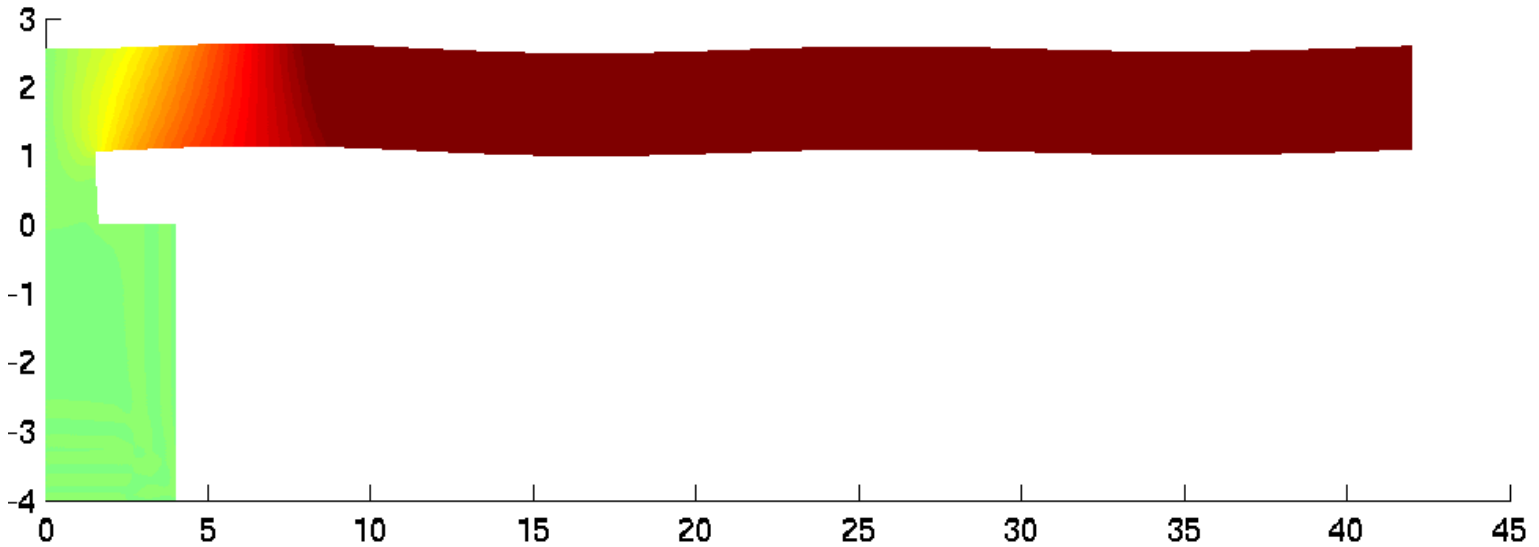
Model reduction performance



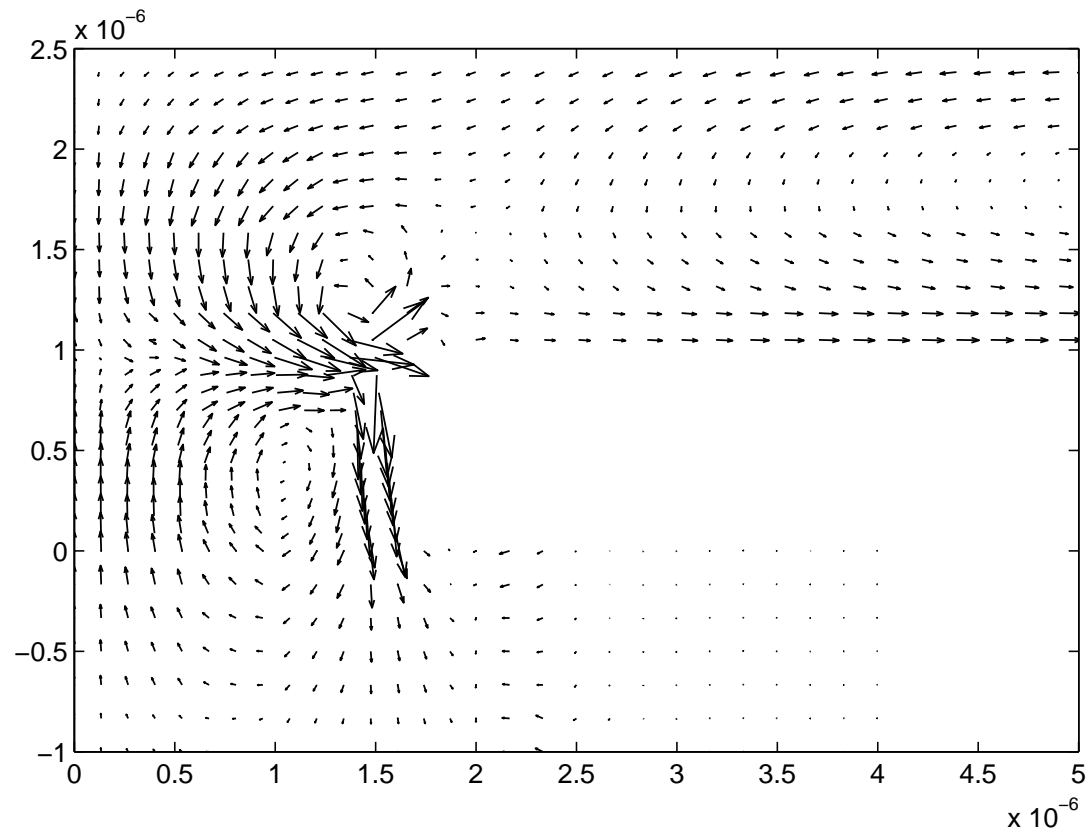
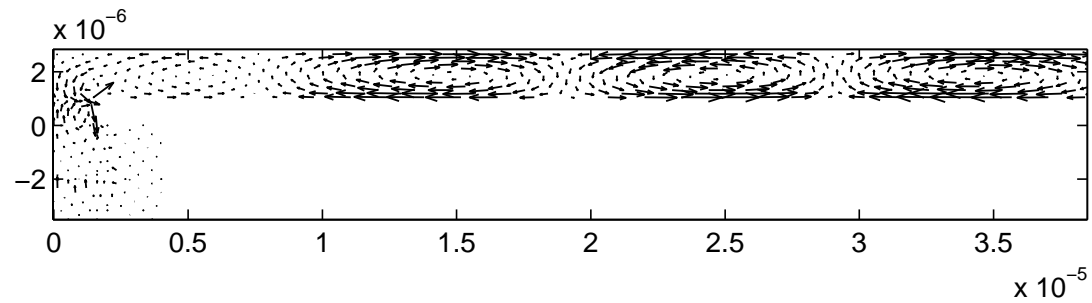
Model reduction performance



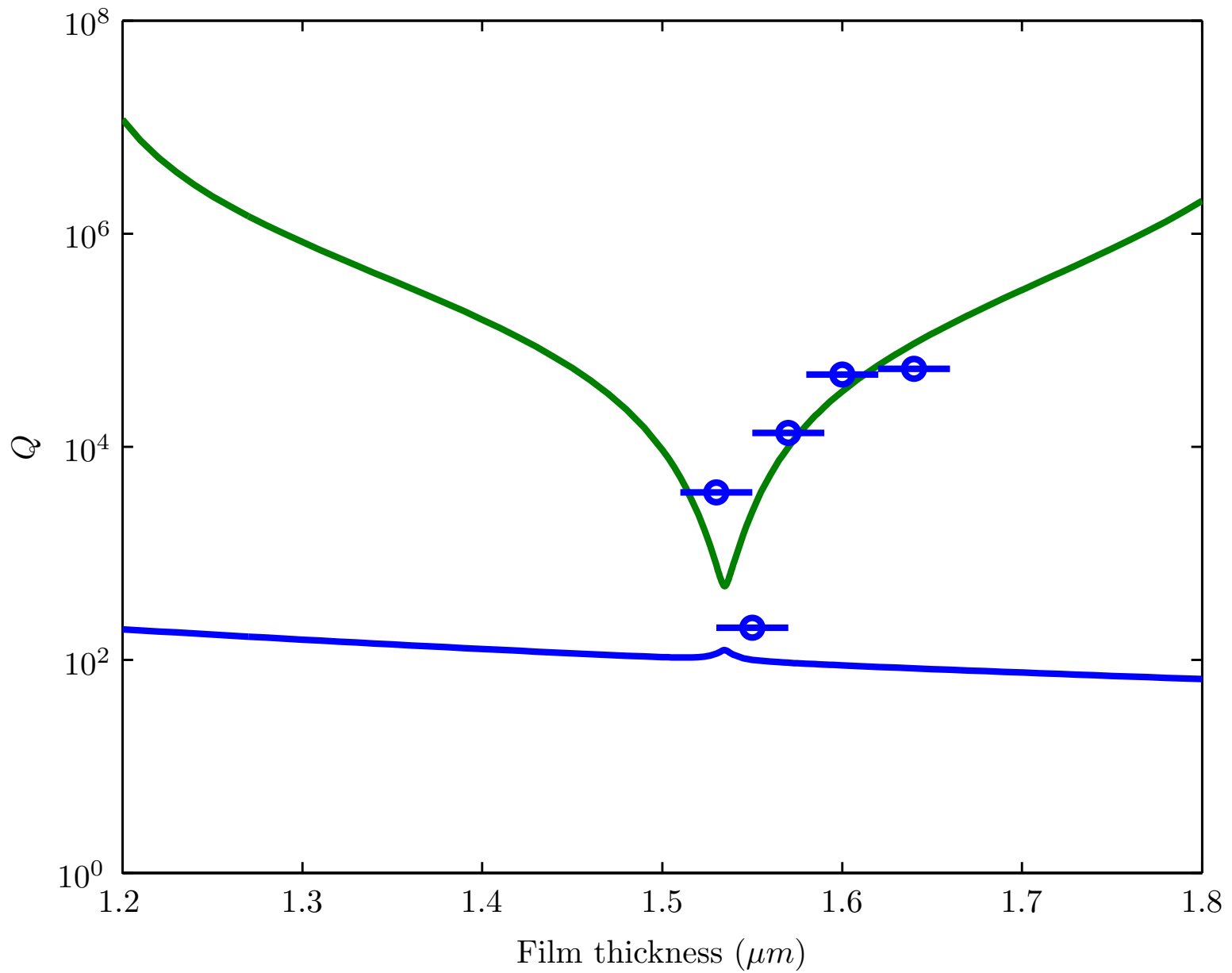
Response of the disk resonator



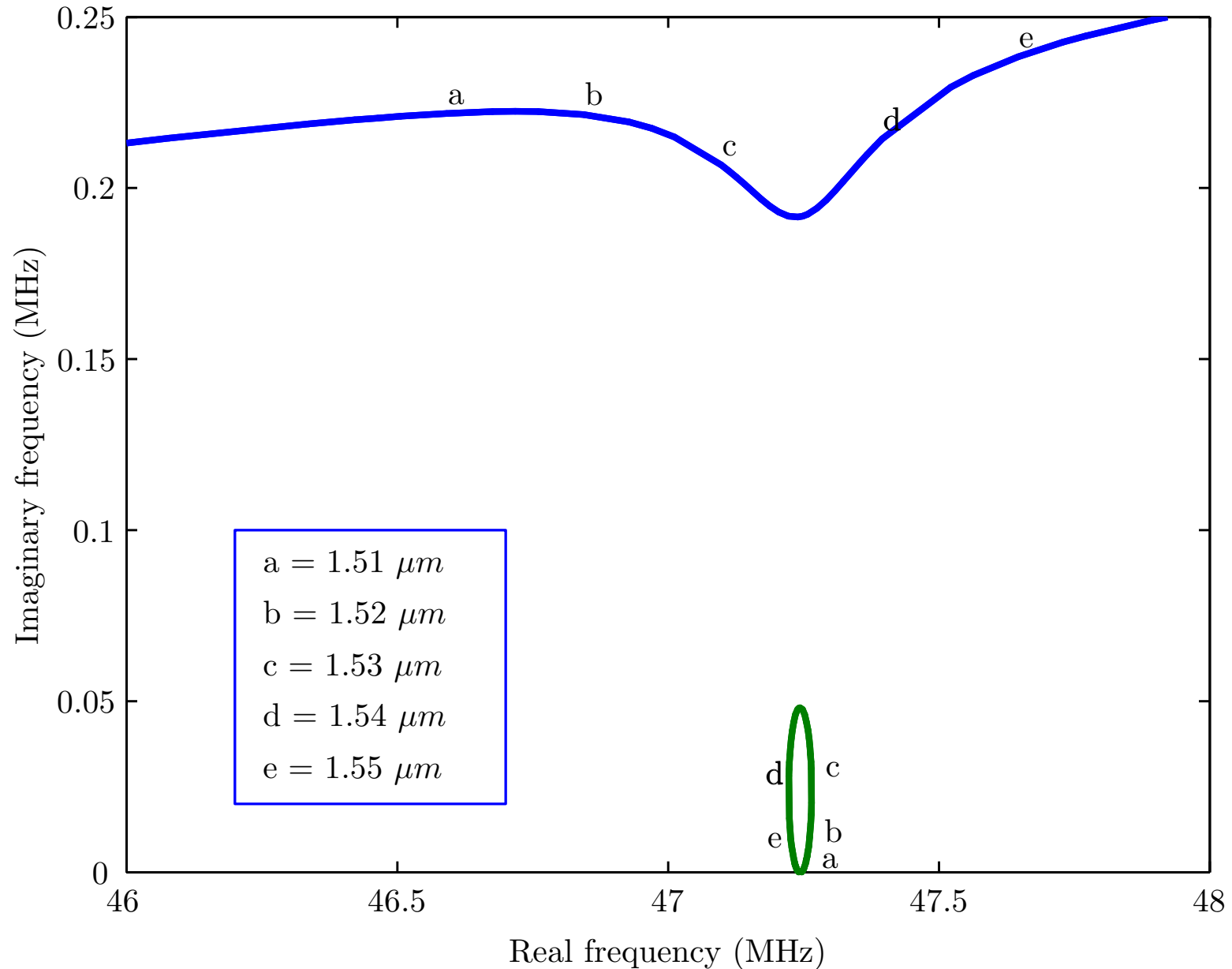
Time-averaged energy flux



Q variation



Explanation of Q variation



Conclusions

- MEMS damping is important and non-trivial
- Elastic PMLs work well for modeling anchor loss
 - Formulation fits naturally with mapped elements
 - Estimate multi-D performance with simple models
- Use complex symmetry to compute eigenvalues and reduced models
- Simulations show effects that hand analysis misses

Reference:

Bindel and Govindjee, “Elastic PMLs for resonator anchor loss simulation,” *IJNME* (to appear).